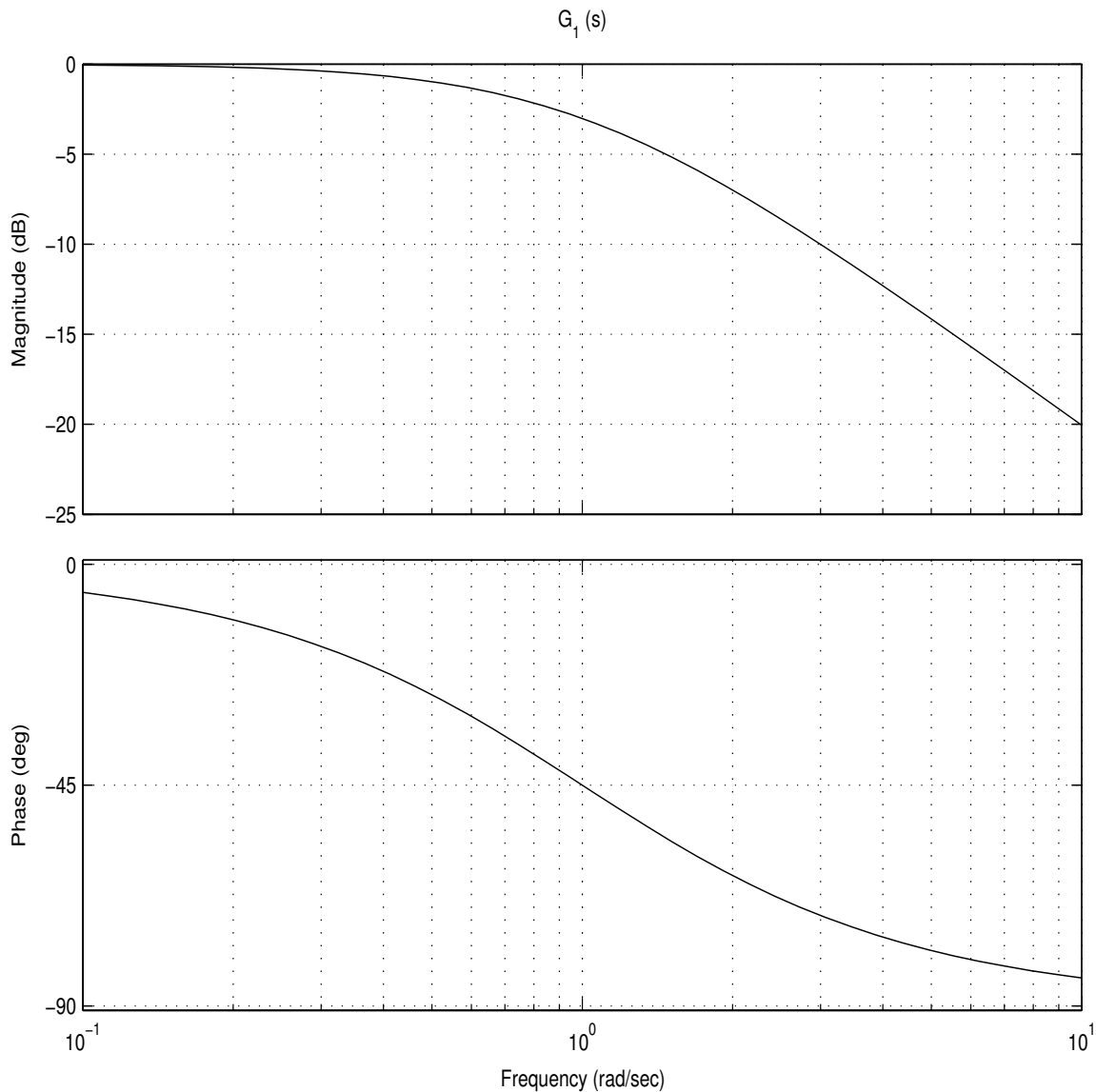


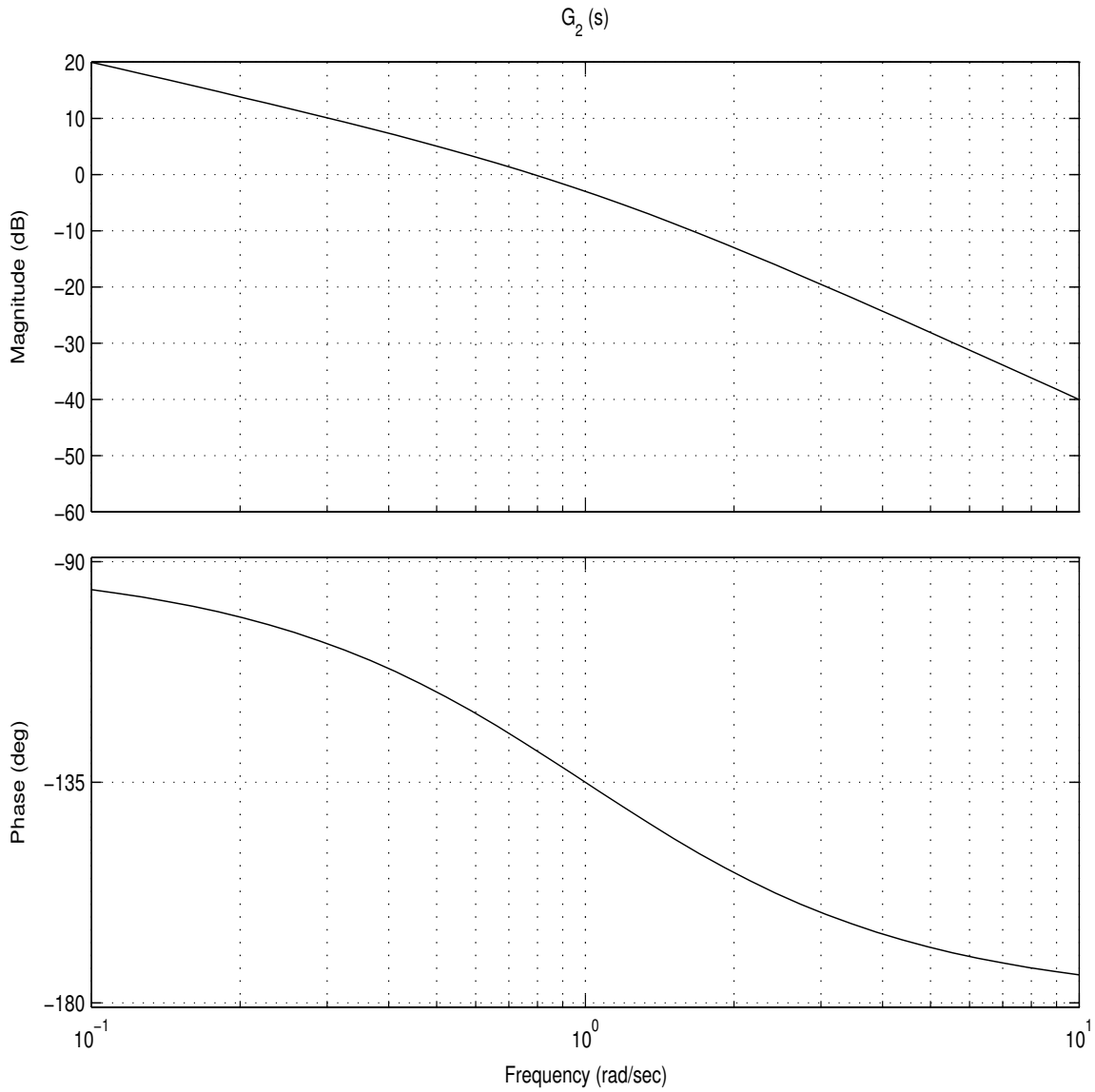
PROBLEM SET 10
Solutions

Problem 1

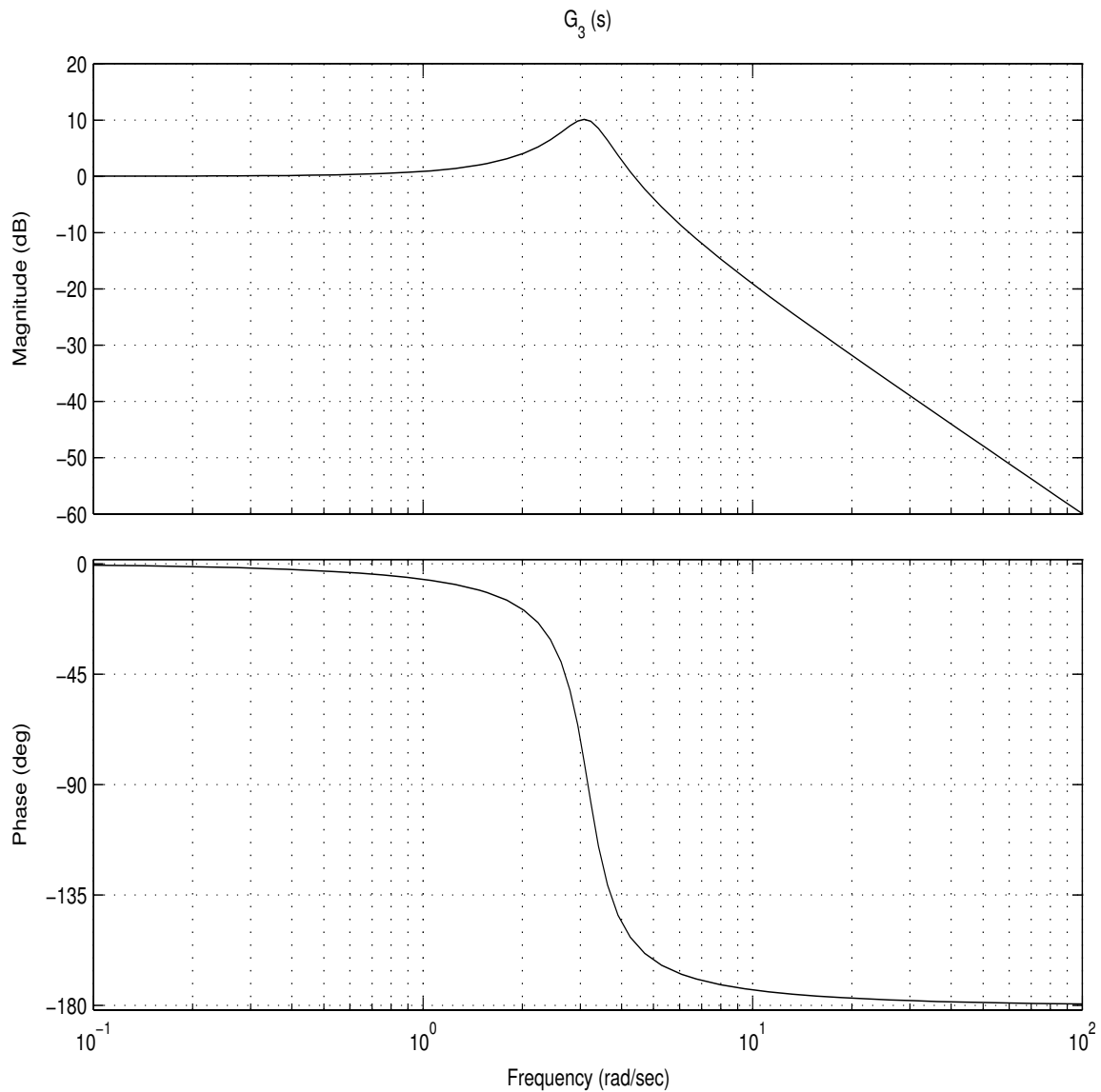
- The Bode gain is $K_{Bode} = 1$, and there are no integrators, so the low-frequency asymptote is $M_{dB} = 0$. Again because there are no integrators, the phase at low frequencies is 0° .
 - The break frequency of the pole is $\omega = 1$. At this frequency, the slope of the magnitude plot changes by -20 dB/dec. The phase changes by -5° one decade before the break frequency (at $\omega = 0.1$), by -45° at the break frequency (at $\omega = 1$), and by -85° one decade after the break frequency (at $\omega = 10$). At high frequencies, the phase approaches -90° .



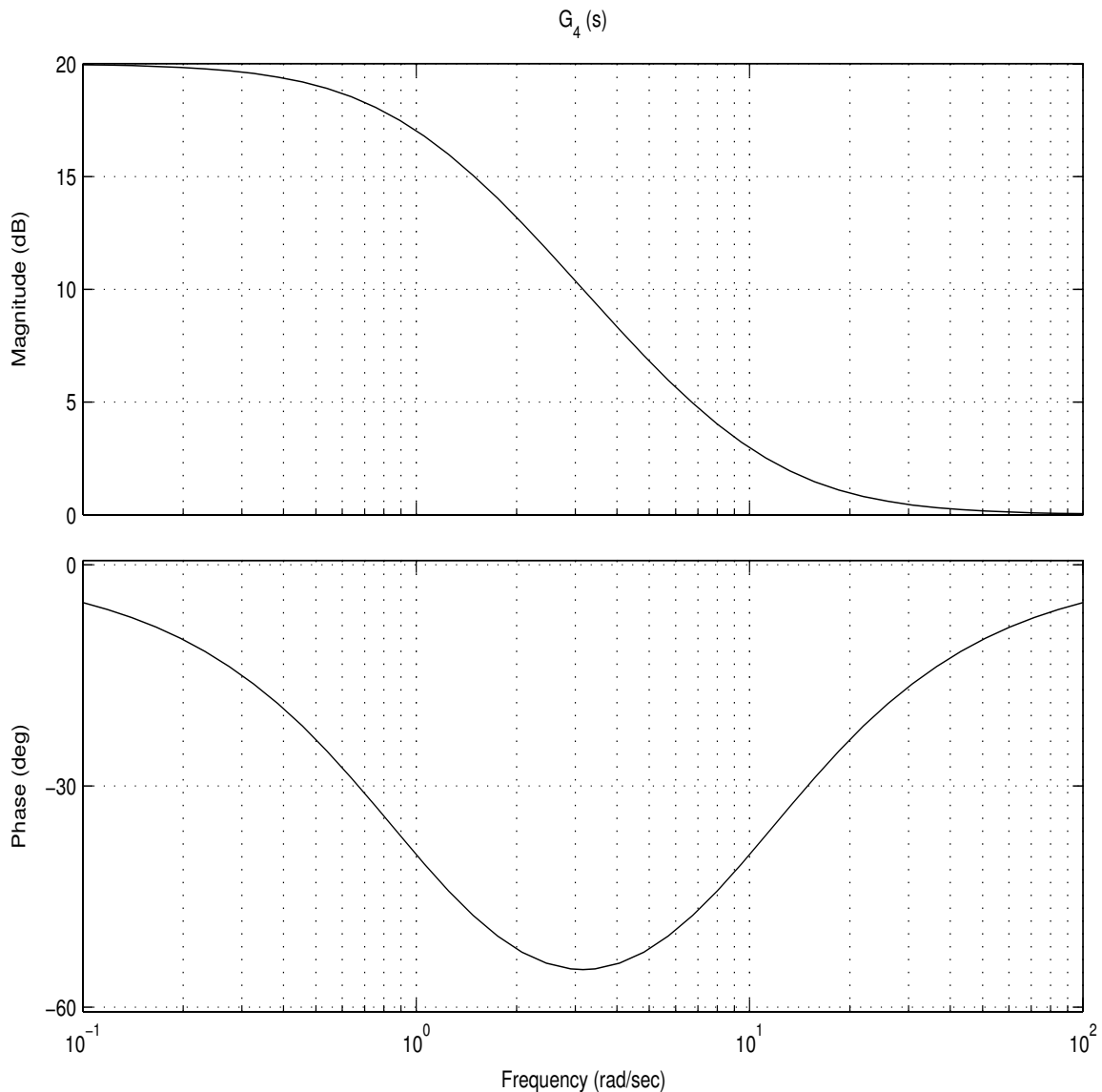
- 2.
- $K_{Bode} = 1$ and there is one integrator, so the low-frequency asymptote has a slope of -20 dB/dec and a magnitude of 1 ($= 0$ dB) at $\omega = 1$. Because there is one integrator, the phase at low frequencies is -90° .
 - The pole at $s = -1$ breaks at $\omega = 1$, so the slope of the magnitude plot changes by -20 dB/dec at that point. (It goes from -20 dB/dec to -40 dB/dec.) The phase is approximately -95° at $\omega = 0.1$, -135° at $\omega = 1$, and -175° at $\omega = 10$. The phase at high frequencies approaches -180° .



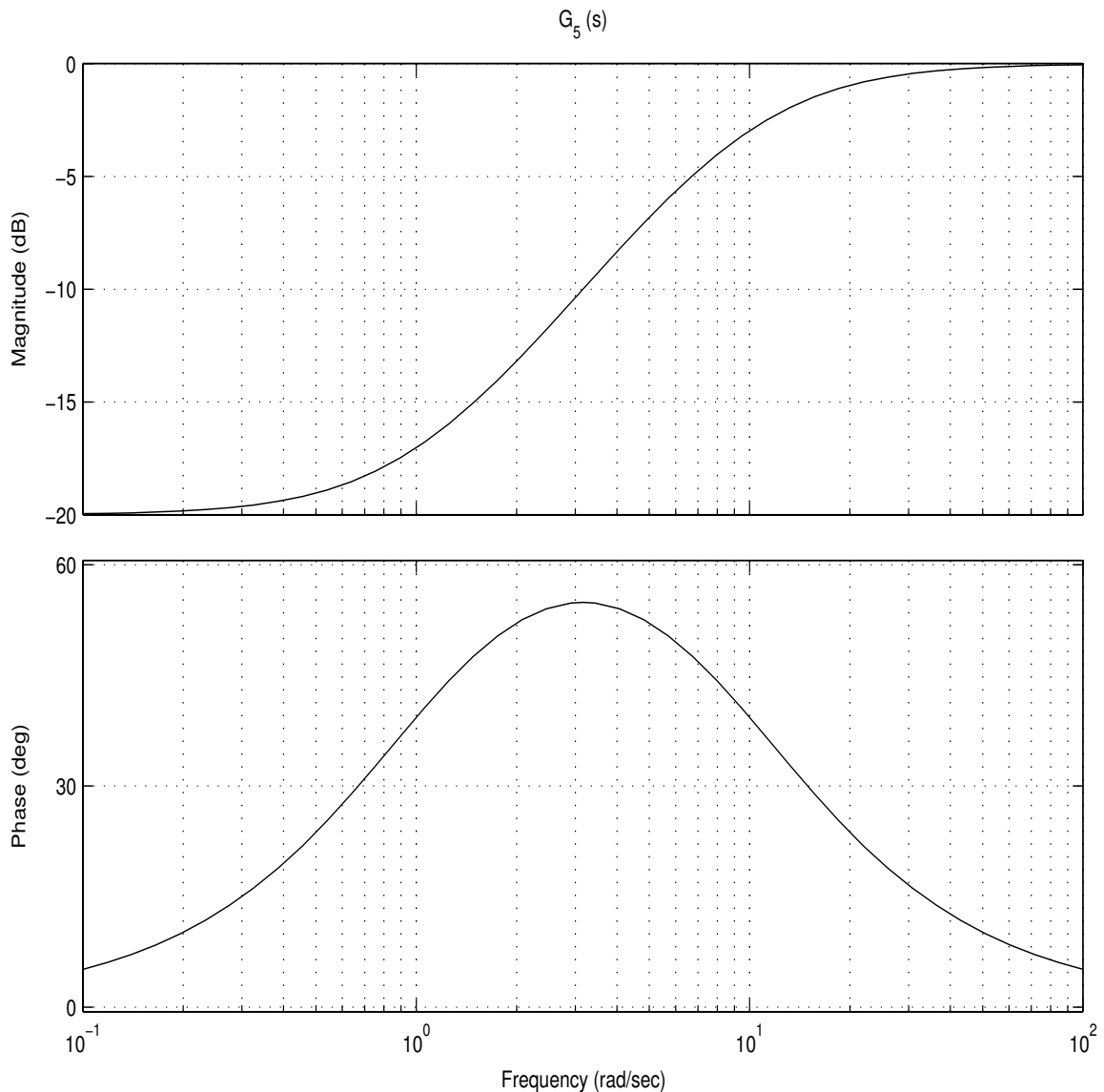
- 3.
- $K_{Bode} = 1$ and there are no integrators, so the low-frequency asymptote is $M_{dB} = 0$ and the phase at low frequencies is zero.
 - There is a complex conjugate set of poles given by $s^2 + 2\zeta\omega_n s + \omega_n^2 = s^2 + s + 10$. So $\omega_n = \sqrt{10}$ and $\zeta = \frac{1}{2\sqrt{10}} \approx 0.16$.
 - The break frequency for the complex conjugate set of poles is at $\omega = \omega_n = \sqrt{10}$ (half-way between $\omega = 1$ and $\omega = 10$ on a log scale). At this frequency, the slope of the magnitude plot changes by -40 dB/dec. The phase changes by -180° between approximately one decade before the break frequency ($\omega = 1/\sqrt{10}$) and one decade after ($\omega = 10\sqrt{10}$). (The steepness of the phase change depends on the value of ζ , but this approximation is good enough for this class.)
 - From Figure 7.13 in Van De Vegte, a damping ratio of 0.16 corresponds to a peak in the magnitude plot of approximately 10 dB at $\omega = \omega_n$. The plot below (from Matlab) shows this peak.



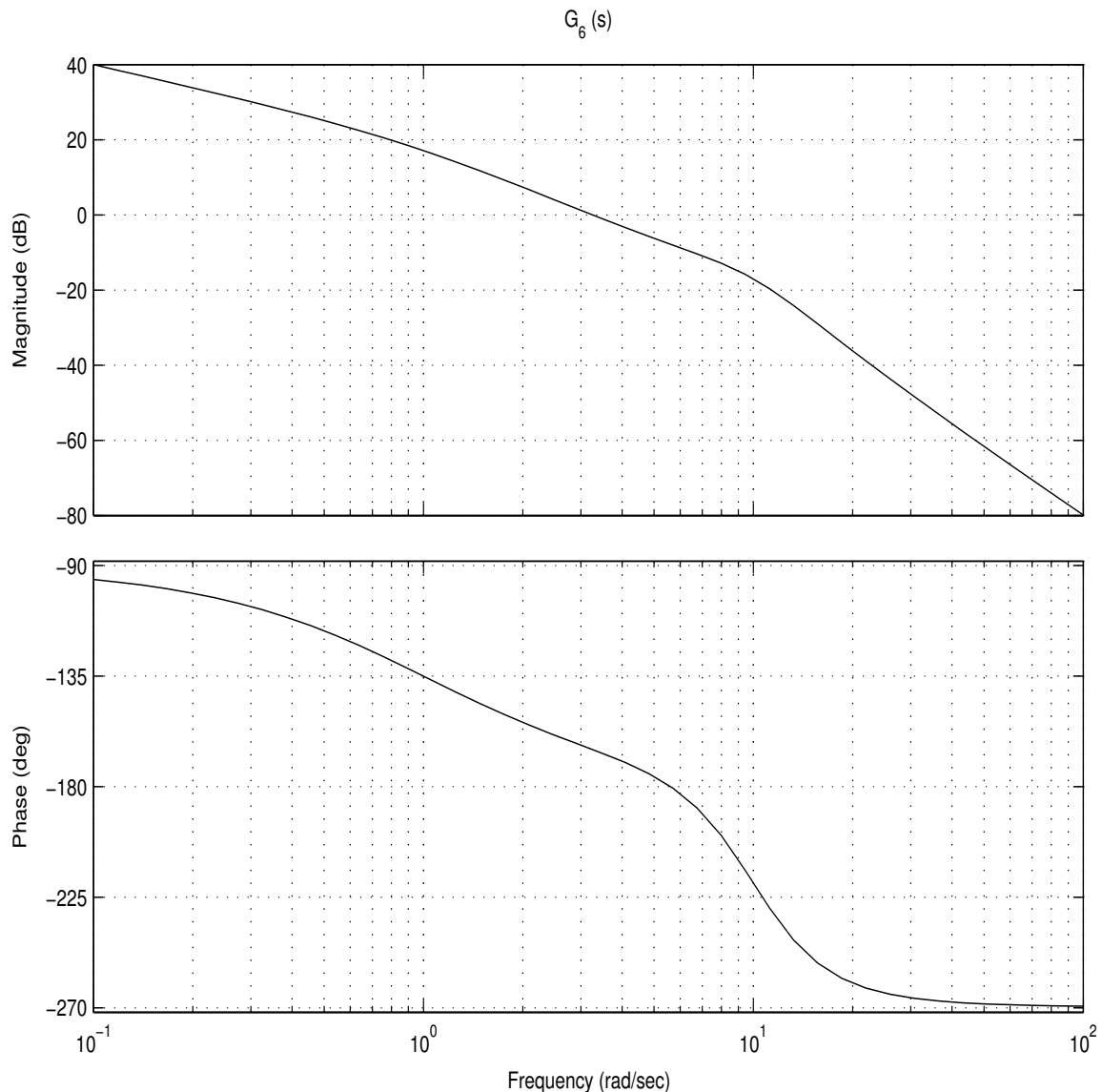
- 4.
- $K_{Bode} = 10$ and there are no integrators, so the low-frequency asymptote is $M_{dB} = 20$ and the phase at low frequencies is zero.
 - The first break frequency is due to the pole at $s = 1$. So at $\omega = 1$ the slope of the magnitude plot changes from 0 dB/dec to -20 dB/dec. Then the break frequency due to the zero is at $\omega = 10$, so at that frequency the slope changes from -20 dB/dec back to 0 dB/dec.
 - At $\omega = 0.1$, the pole contributes -5° and the zero contributes nothing, so the phase is -5° .
 - At $\omega = 1$, the pole contributes -45° and the zero contributes $+5^\circ$, so the phase is -40° .
 - At $\omega = 10$, the pole contributes -85° and the zero contributes $+45^\circ$, so the phase is -40° .
 - At $\omega = 100$, the pole contributes -90° and the zero contributes $+85^\circ$, so the phase is -5° .
 - As $\omega \rightarrow \infty$, the phase contribution from the pole and zero cancel and so the phase is 0° .



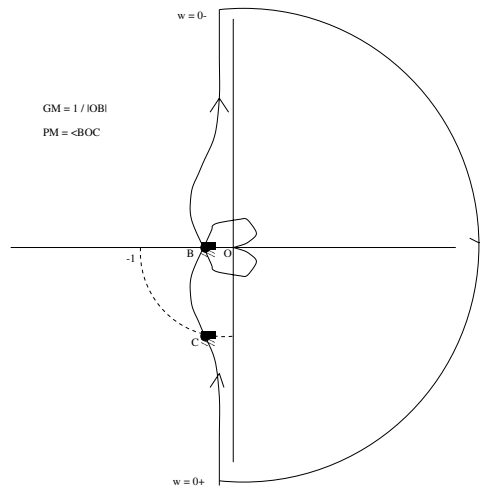
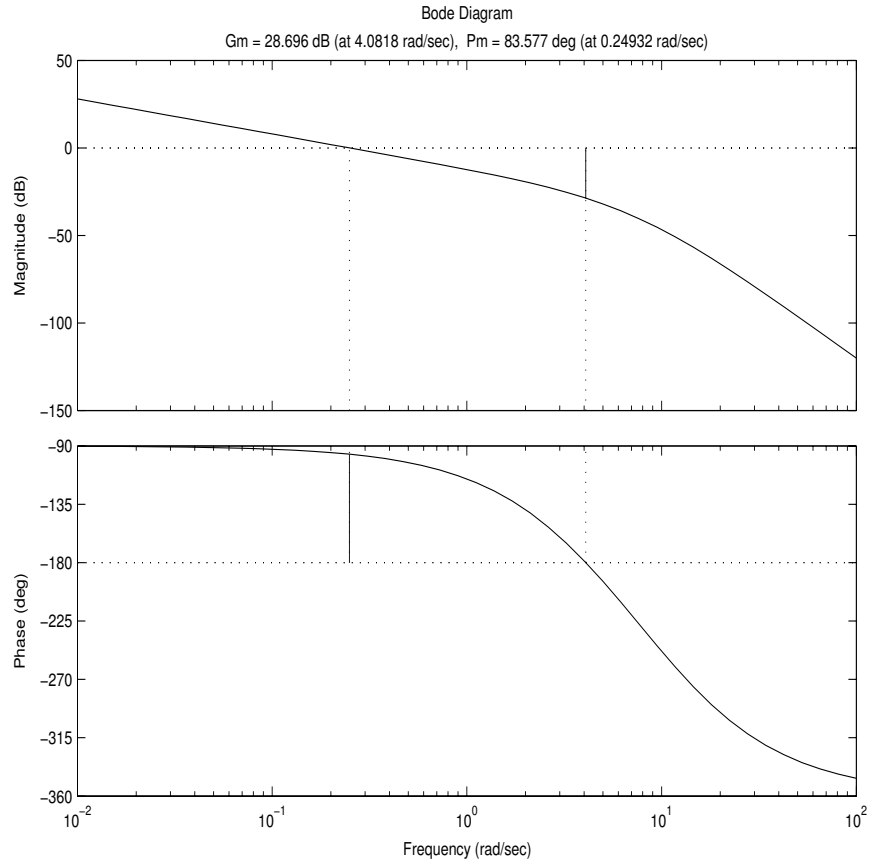
- 5.
- $K_{Bode} = 0.1$ and there are no integrators, so the low-frequency asymptote is $M_{dB} = -20$ and the phase at low frequencies is zero.
 - The first break frequency is due to the zero at $s = 1$. So at $\omega = 1$ the slope of the magnitude plot changes from 0 dB/dec to +20 dB/dec. Then the break frequency due to the pole is at $\omega = 10$, so at that frequency the slope changes from +20 dB/dec back to 0 dB/dec.
 - At $\omega = 0.1$, the zero contributes $+5^\circ$ and the pole contributes nothing, so the phase is $+5^\circ$.
 - At $\omega = 1$, the zero contributes 45° and the pole contributes -5° , so the phase is $+40^\circ$.
 - At $\omega = 10$, the zero contributes 85° and the pole contributes -45° , so the phase is $+40^\circ$.
 - At $\omega = 100$, the zero contributes 90° and the pole contributes -85° , so the phase is $+5^\circ$.
 - As $\omega \rightarrow \infty$, the phase contribution from the pole and zero cancel and so the phase is 0° .



- 6.
- $K_{Bode} = 10$ and there is one integrator, so the low-frequency asymptote has a slope of -20 dB/dec and a magnitude of 20 dB at $\omega = 1$. Because there is one integrator, the phase at low frequencies is -90° .
 - The complex conjugate poles have $\omega_n = \sqrt{100} = 10$ and $\zeta = 0.5$.
 - The pole at $s = -1$ breaks at $\omega = 1$, so the slope of the magnitude plot changes from -20 dB/dec to -40 dB/dec at that frequency. The pole contributes -5° of phase shift at $\omega = 0.1$, -45° at $\omega = 1$, and -85° at $\omega = 10$.
 - The zero at $s = -10$ and the complex conjugate poles both break at $\omega = 10$. The zero changes the slope by +20 dB/dec, and the complex poles change it by -40 dB/dec, so the net effect is a change in slope of -20 dB/dec. The slope therefore changes from -40 dB/dec to -60 dB/dec.
 - The zero contributes $+5^\circ$ of phase shift at $\omega = 1$, $+45^\circ$ at $\omega = 10$, and $+85^\circ$ at $\omega = 100$.
 - The complex poles contribute approximately -5° of phase shift at $\omega = 1$, -90° at $\omega = 10$, and -175° at $\omega = 100$. (If you wanted to plot the phase more accurately than that, you would use Matlab!)
 - Because the damping ratio is so high, there is almost no peak in the magnitude at $\omega = \omega_n$.
 - There are three more poles than zeros, so as $\omega \rightarrow \infty$, the phase approaches -270° .



Problem 2



Problem 3

1. The first thing to notice is that the low-frequency asymptote is $M_{dB} = 20$ dB and the phase at low frequencies is zero, so there are no integrators and the Bode (standard) gain of $G_1(s)$ is 10. At around $\omega = 1$, the slope of the magnitude plot drops to -20 dB/dec and the phase turns negative, so there is a pole at $s = -1$. Then at around $\omega = 30$, the slope of the magnitude plot reverts to zero, and the phase starts getting more positive, so there is a zero at $s = -30$. Finally, at $\omega = 1000$, the slope drops to -20 dB/dec again and the phase gets more negative, so there is a pole at $s = -1000$. The final transfer function is then:

$$G_1(s) = \frac{10\left(\frac{s}{30} + 1\right)}{(s + 1)\left(\frac{s}{1000} + 1\right)}$$

Check: there is one more pole than zero, so the phase at very high frequencies should be -90° (it is).

2. Now the low-frequency asymptote has a slope of -20 dB/dec and a magnitude of 0 dB at $\omega = 1$, and the phase at very low frequencies is -90° . So there is one integrator, and the Bode gain is 1.

At $\omega = 1$, the slope changes to 0 dB/dec, indicating the presence of a zero. However, the phase changes by -90° , so the zero must be in the right half-plane at $s = 1$.

Then at $\omega = 100$, the slope changes to -20 dB/dec, indicating the presence of a pole. However, the phase changes by $+90^\circ$ so the pole must be in the right half-plane at $s = 100$.

$$\Rightarrow G_2(s) = \frac{-s + 1}{s\left(-\frac{s}{100} + 1\right)}$$

Problem 4

1. The crossover frequency is the frequency at which $|G(j\omega)| = 1$ or 0 dB. The gain margin is the difference in magnitude between 0 dB and $|G(j\omega)|$ when the phase is -180° . The phase margin is the difference in phase between -180° and $\angle G(j\omega)$ when the magnitude is 1. From the tabulated data (or the Bode plot), we get:

$$\begin{aligned}\omega_c &\approx 3.1 \\ GM &\approx \frac{1}{0.25} = 4 \\ \phi_m &\approx 180^\circ - 135^\circ = 45^\circ\end{aligned}$$

2. From Figure 8.7: a phase margin of 45° corresponds to a damping ratio of $\zeta \approx 0.43$. From Figure 8.6: this damping ratio corresponds to $\frac{\omega_c}{\omega_n} \approx 0.84$, so $\omega_n = \frac{3.1}{0.84} = 3.7$.
3. From Table 8.4.1 (or Figure 8.7): to get $\zeta = 0.6$, we need $\phi_m = 59^\circ$.
4. To get this phase margin, we want the phase of $G(j\omega)$ to be $-180^\circ + 59^\circ = -121^\circ$ when $|G(j\omega)| = 1$. From the tabulated data, $G(j\omega)$ has this phase at $\omega = 2.46$ and $|G(j\omega)| = 1.42$ at this frequency. So we need to multiply $G(j\omega)$ by a factor of $1/1.42 = .704$ to get the desired phase margin. So $K = .704$.
5. From Table 8.4.1: $\zeta = 0.6 \Rightarrow \frac{\omega_c}{\omega_n} = 0.72$. The new crossover frequency is 2.46, so the new natural frequency is $\omega_n = \frac{2.46}{0.72} = 3.4$.