2.161 Signal Processing: Continuous and Discrete Fall 2008

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MASSACHUSETTS INSTITUTE OF TECHNOLOGY DEPARTMENT OF MECHANICAL ENGINEERING

2.161 Signal Processing - Continuous and Discrete Fall Term, 2007

Quiz #1 — October 18, 2007

Notes:

- The quiz is open-book.
- The time allowed is ninety minutes.
- There are six problems, answer them all.
- Partial credit will be given.

<u>Problem 1:</u> (15 points)

Show that to interpolate a value to the mid-point of a sampling interval in the sample set $\{f_n\}, n = 0 \dots N - 1$, where $f_n = f(n\Delta T)$, the cardinal (Whittaker) reconstructor may be written

$$f(n\Delta T + \Delta T/2) = \frac{1}{\pi} \sum_{k=0}^{N-1} f_k \frac{(-1)^k}{(n-k+1/2)}$$

Problem 2: (20 points)

- (a) Design (find the transfer function of) a unity-gain 3rd-order low-pass Butterworth analog filter with a -3dB cut-off frequency of 10 rad/s.
- (b) Make a pole-zero plot for your filter.
- (c) Convert your design to a high-pass filter with the same cut-off frequency.
- (d) If (do not do it) you were to convert your design to a band-pass filter, what would be the order of the filter, how many zeros would be created, and where in the *s*-plane would the zeros lie?

Problem 3: (15 points)

Many real-life signal processing problems involve waveforms containing echoes, or reverberation. Consider a continuous-time linear filter with impulse response

$$h(t) = \delta(t) + \delta(t - \tau)$$

Assume that the filter is excited with waveform f(t), and that the output is y(t)

(i) Show that the output contains an echo, that is

$$y(t) = f(t) + f(t - \tau)$$

(ii) Find the frequency response H(jω) for the filter, and express it in terms of its magnitude |H(jω)|, and phase ∠H(jω).
Hint:

$$1 + e^{-j\theta} = e^{-j\theta/2} \left[e^{j\theta/2} + e^{-j\theta/2} \right]$$

(iii) Assume $f(t) = \sin \omega t$. At what values of ω will the filter exhibit no steady-state output?

<u>Problem 4:</u> (20 points)

An AM (amplitude-modulated) radio signal $f_{AM}(t)$ is described by

$$f_{AM}(t) = (1 + a f_{audio}(t)) \sin(\omega_c t)$$

where $f_{audio}(t)$ is the audio signal, $\sin(\omega_c t)$ is known as radio-frequency *carrier* signal ($f_c = 500 - 1600$ kHz - the AM band), and a is a positive constant that determines the *modulation depth*. (Note that we require $|af_{audio}(t)| < 1$ otherwise we have over-modulation.) the following figure shows an AM signal with an "audio" waveform that is a simple low frequency sinusoid. You can see how the audio signal "modulates" the amplitude of the rf signal.



- (a) Sketch the magnitude of the Fourier transform of $f_{AM}(t)$ when $f_{audio}(t) = 0$.
- (b) Let a = 0.5, and sketch the magnitude of the Fourier transform of $f_{AM}(t)$ when

$$f_{audio}(t) = 0.5\cos(2\pi \cdot 1000t) + 0.25\cos(2\pi \cdot 2000t)$$

(**Hint:** There is no need to actually compute the FT. Consider expanding $f_{AM}(t)$, or simply use properties of the FT.)

(c) Use your result from (b) to generalize, and sketch the magnitude spectrum of $f_{AM}(t)$ when $f_{audio}(t)$ has a spectrum (again let a = 0.5):



(d) If f_{audio} has a bandwidth $B = \omega_u - \omega_l$, what is the bandwidth of a band-pass filter that would be necessary to select the signal $f_{AM}(t)$ out of all the other AM radio stations?

<u>Problem 5:</u> (15 points)

The most commonly used (approximate) data reconstructor is the "zero-order hold" (ZOH) which simply "holds" the output value $y(n\Delta T)$ as a constant over the reconstruction interval:

$$y(t) = f(n\Delta T)$$
 $n\Delta T \le t < (n+1)\Delta T$

where $f(n\Delta T)$ is the value of input at time $t = n\Delta T$. The output of the ZOH therefore looks like a staircase function, with discrete jumps to a new value at each update time $n\Delta T$:



(a) Show that the ZOH can be represented as a linear filter with a pulse-like impulse response

$$h(t) = 1 \qquad 0 \le t < \Delta T$$
$$= 0 \qquad \text{otherwise}$$

Hint: Consider the sampled input data waveform $f^*(t)$ as a weighted impulse train at intervals ΔT ,

$$f^*(t) = \sum_{n = -\infty}^{\infty} f(nT)\delta(t - n\Delta T).$$

- (b) Find and sketch the frequency response function $H(j\omega)$ for the ZOH data reconstructor.
- (c) Compare $H(j\omega)$ with the frequency response of the ideal (cardinal) data reconstructor, and comment on why the ZOH is a non-ideal reconstructor.
- (d) The ZOH reconstruction (see the fig above) seems to be slightly delayed from f(t). Determine the delay.

<u>Problem 6:</u> (15 points)

The significant frequency range of an analog signal extends to 10 kHz. Beyond 10 kHz the signal spectrum rolls-off (attenuates) at a rate of 20 dB/decade.

The signal is to be sampled at a rate of 200 kHz. The aliased frequency components introduced into the 10 kHz range of interest must be kept below -60 dB as compared to signal components.

Suppose we use an analog low-pass pre-aliasing filter whose passband is flat over the 10 kHz band, and then attenuates at a rate steep enough to satisfy the above sampling requirements. What is this attenuation rate in dB/decade? What would be the minimum order of a low-pass filter to satisfy this condition?

θ	$\sin \theta$	$\cos \theta$	an heta
0	0	1	0
$\pi/6$	1/2	$\sqrt{3}/2$	$\sqrt{3}/3$
$\pi/4$	$\sqrt{2}/2$	$\sqrt{2}/2$	1
$\pi/3$	$\sqrt{3}/2$	1/2	$\sqrt{3}$
$\pi/2$	1	0	∞

x	$\log_e(x)$	$\log_{10}(x)$
2	0.6931	0.3010
3	1.0986	0.4771
5	1.6094	0.6990
10	2.3026	1.0000

$\sin(a+b)$	=	$\sin(a)\cos(b) + \cos(a)\sin(b)$
$\sin(a-b)$	=	$\sin(a)\cos(b) - \cos(a)\sin(b)$
$\cos(a+b)$	=	$\cos(a)\cos(b) - \cos(a)\sin(b)$
$\cos(a-b)$	=	$\cos(a)\cos(b) + \cos(a)\sin(b)$
$\tan(a+b)$	=	$\frac{\tan(a) + \tan(b)}{1 - \tan(a)\tan(b)}$
$\tan(a-b)$	=	$\frac{\tan(a) - \tan(b)}{1 + \tan(a)\tan(b)}$

Some useful/useless information: