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2.830J / 6.780J / ESD.63J Control of Manufacturing Processes (SMA 6303)  
Spring 2008

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# Control of Manufacturing Processes

Subject 2.830

Spring 2007

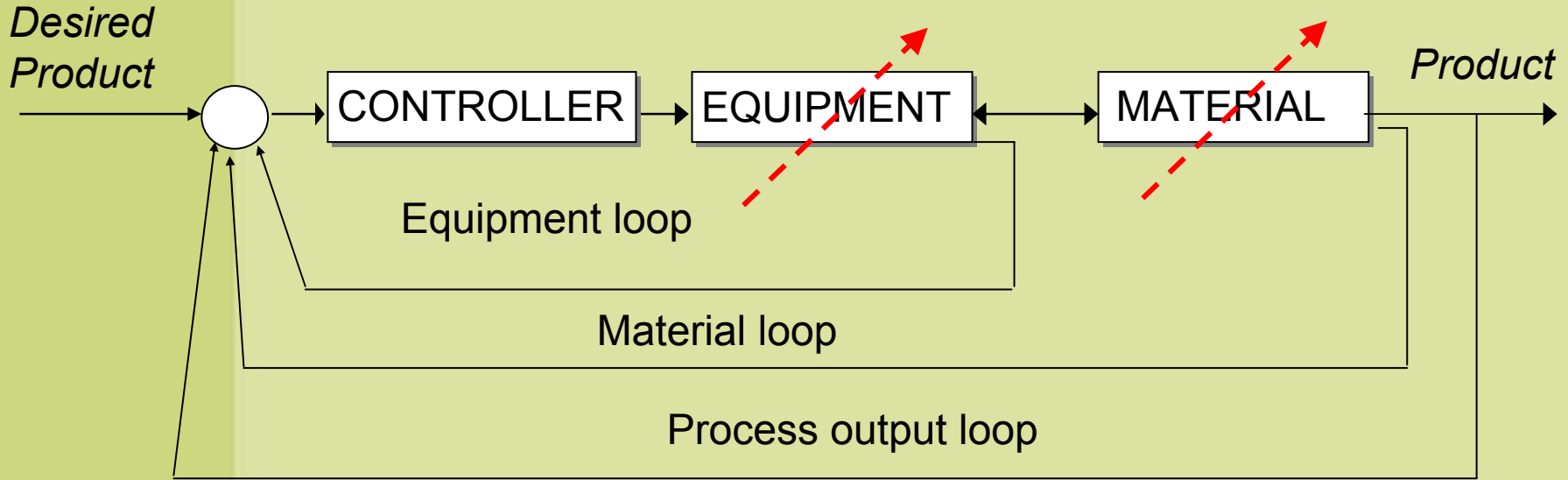
Lecture #20

“Cycle To Cycle Control:  
The Case for using Feedback and SPC”

May 1, 2008



# The General Process Control Problem



## Control of **Equipment:**

Forces,  
 Velocities  
 Temperatures,  
 , ..

## Control of **Material**

Strains  
 Stresses  
 Temperatures,  
 Pressures, ..

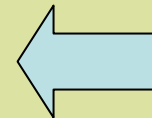
## Control of **Product:**

Geometry  
 and  
 Properties

# Output Feedback Control

$$\Delta Y = \frac{\partial Y}{\partial \alpha} \Delta \alpha + \frac{\partial Y}{\partial u} \Delta u$$

$$\frac{\partial Y}{\partial u} \Delta u = -\frac{\partial Y}{\partial \alpha} \Delta \alpha$$



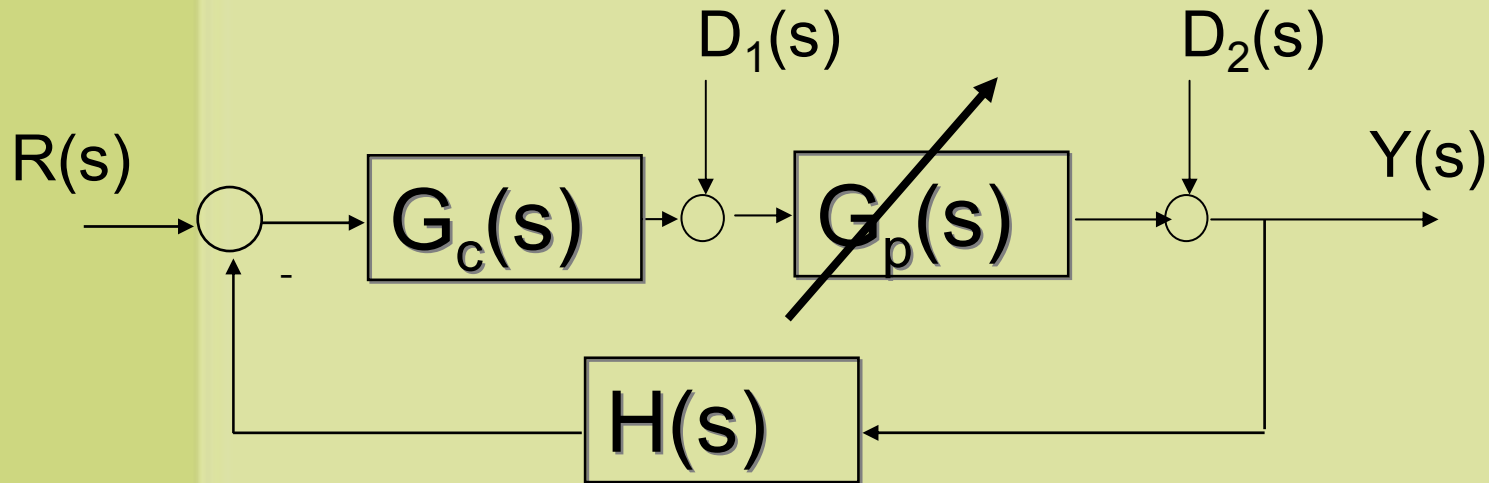
Manipulate  
Actively  
Such that

*Compensate for Disturbances*

# Process Control Hierarchy

- Reduce Disturbances
  - Good Housekeeping
  - Standard Operations (SOP's)
  - **Statistical Analysis and Identification of Sources (SPC)**
  - **Feedback Control of Machines**
- Reduce Sensitivity (increase “Robustness”)
  - **Measure Sensitivities via Designed Experiments**
  - Adjust “free” parameters to minimize
- Measure output and manipulate inputs
  - **Feedback control of Output(s)**

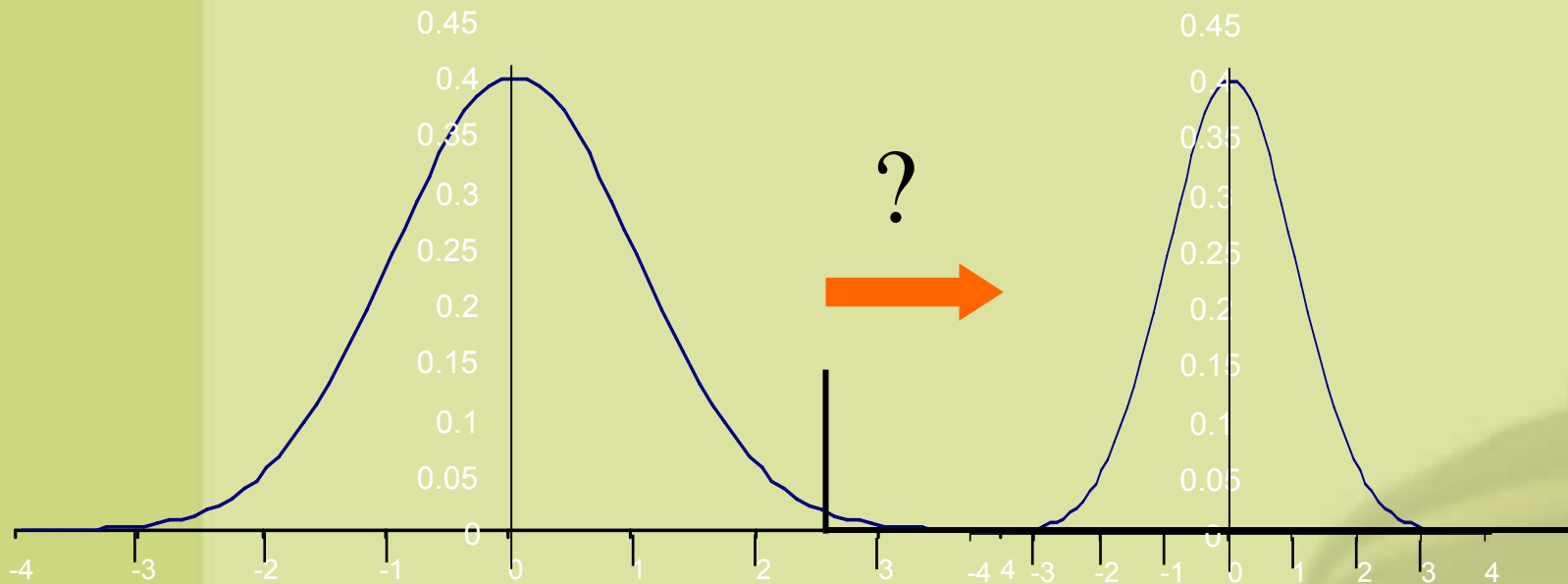
# The Generic Feedback “Regulator” Problem



- Minimize the Effect of the “D’s”
- Minimize Effect of Changes in  $G_p$
- Follow  $R$  exactly

# Effect of Feedback on Random Disturbances

- Feedback Minimizes Mean Shift (Steady-State Component)
- Feedback Can Reduce Dynamic Disturbances



# Typical Disturbances

- Equipment Control
  - External Forces Resisting Motion
  - Environment Changes (e.g Temperature)
  - Power Supply Changes
- Material Control
  - Constitutive Property Changes
    - Hardness
    - Thickness
    - Composition
    - ...



# The Dynamics of Disturbances

- Slowly Varying Quantities
- Cyclic
- Infrequent Stepwise
- Random

# Example: Material Property Changes

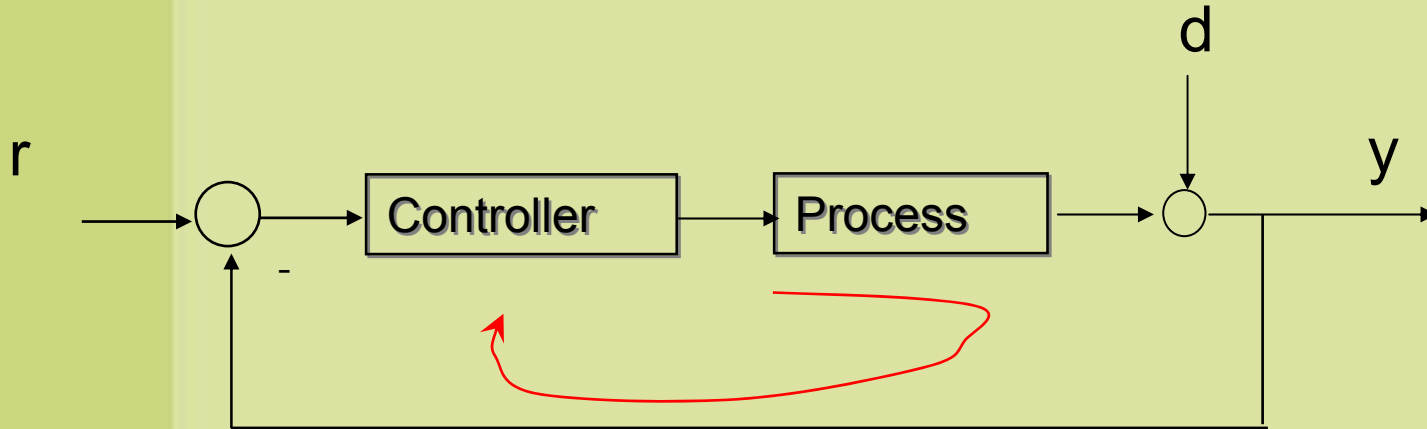
- A constitutive property change from workpiece to workpiece
  - In-Process Effect?
    - A new constant parameter
    - Different outcome each cycle
  - Cycle to Cycle Effect
    - Discrete random outputs over time

# What is Cycle to Cycle?

- Ideal Feedback is the Actual Product Output
- This Measurement Can Always be made After the Cycle
- Equipment Inputs can Always be Adjusted Between Cycles
- Within the Cycle Inputs Are Fixed

# What is Cycle to Cycle?

- Measure and Adjust Once per Cycle



Execute the Loop Once Per Cycle

Discrete Intervals rather than Continuous

# Run by Run Control

- Developed from an SPC Perspective
- Primarily used in Semiconductor Processing
- Similar Results, Different Derivations
- More Limited in Analysis and Extension to Larger problems

Box, G., Luceno, A., “Discrete Proportional-Integral Adjustment and Statistical Process Control,” *Journal of Quality Technology*, vol. 29, no. 3, July 1997. pp. 248-260.

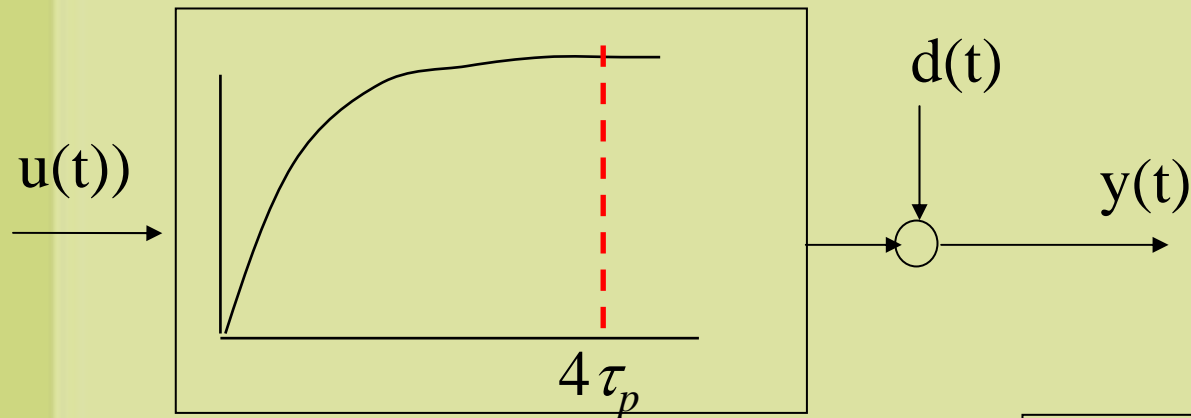
Sachs, E., Hu, A., Ingolfsson, A., “Run by Run Process Control: Combining SPC and Feedback Control.” *IEEE Transactions on Semiconductor Manufacturing*, 1995, vol. 8, no. 1, pp. 26-43.

# Cycle to Cycle Feedback Objectives

- How to Reduce  $E(L(x))$  & Increase  $C_{pk}$  with Feedback?
- Bring Output Closer to Target
  - Minimize Mean or Steady - State Error
- Decrease Variance of Output
  - Reject Time Varying Disturbances

# A Model for Cycle to Cycle Feedback Control

- *Simplest* In-Process Dynamics:

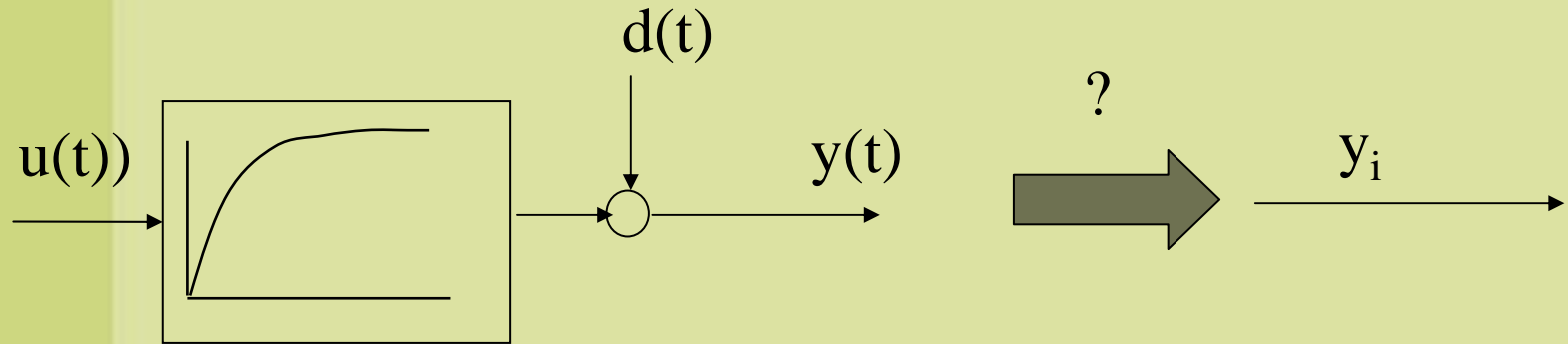


Cycle Time  $T_c \geq 4\tau_p$

$d(t)$  = disturbances seen at the output (e.g. a Gaussian noise)

$\tau_p$  = Equivalent Process Time Constant

# Discrete Product Output Measurement

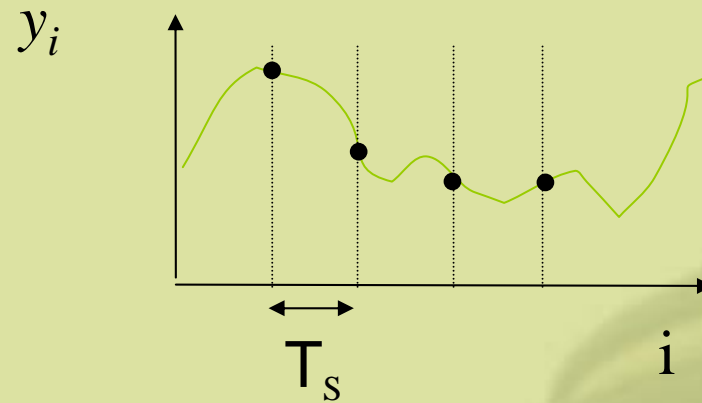
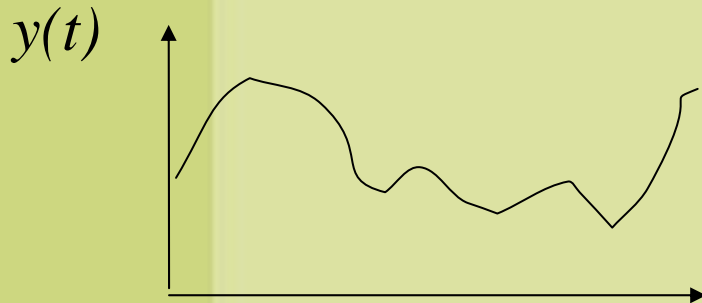
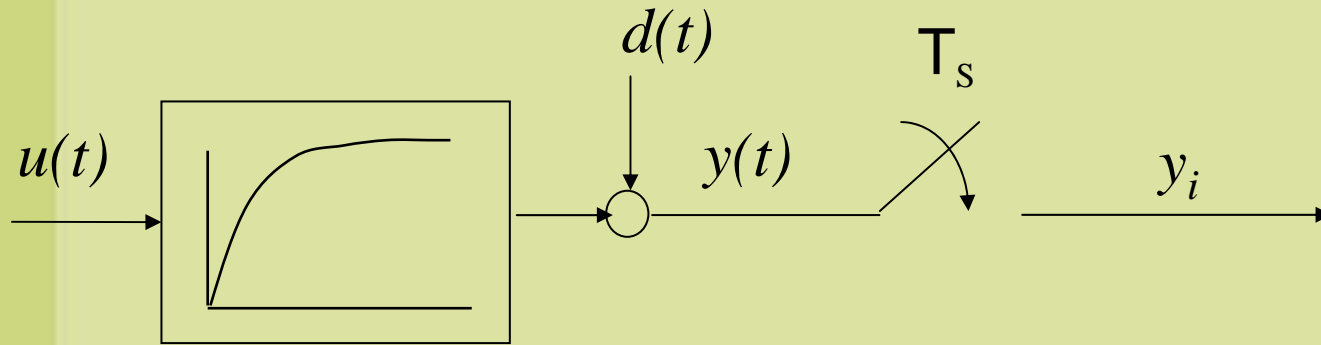


Continuous variable  $y(t)$  to sequential variable  $y_i$

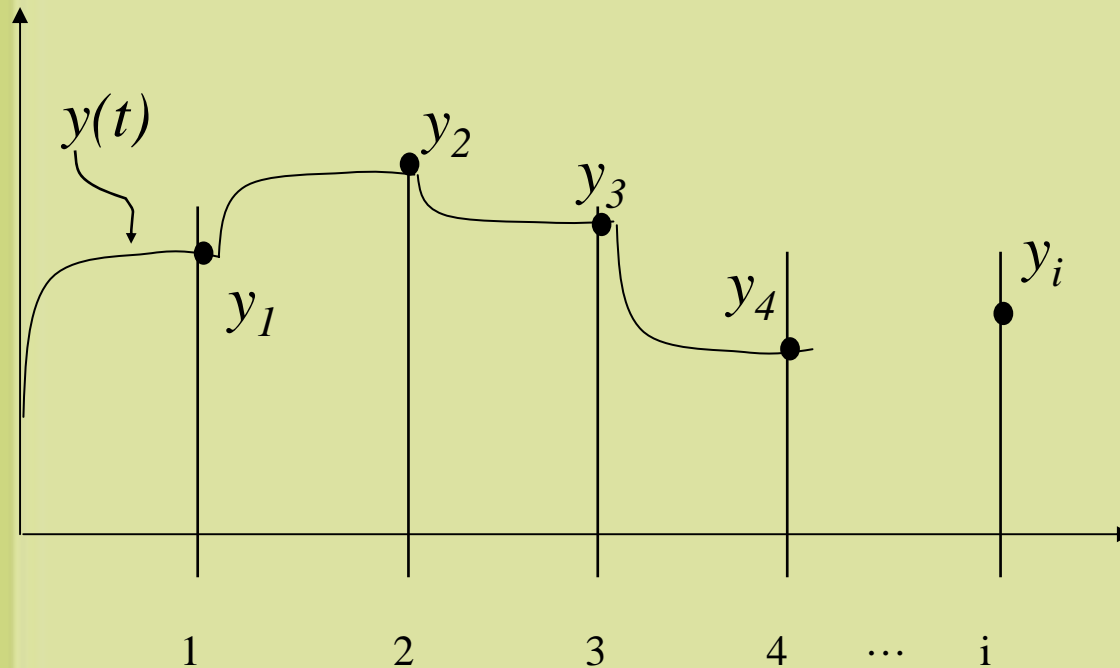
$i$  = time interval or cycle number



# The Sampler

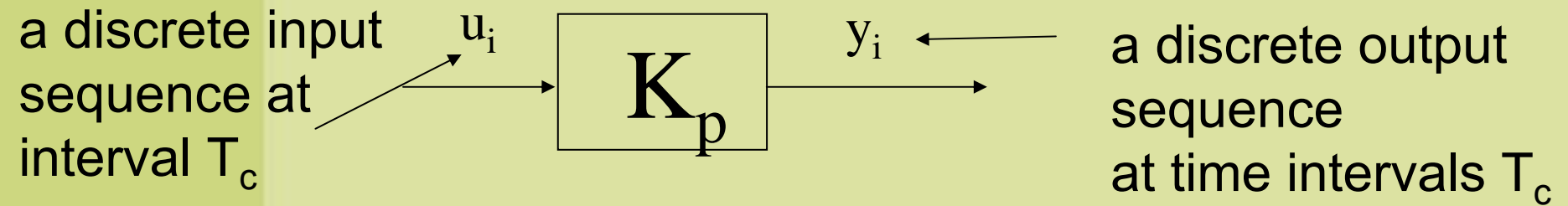


# A Cycle to Cycle Process Model

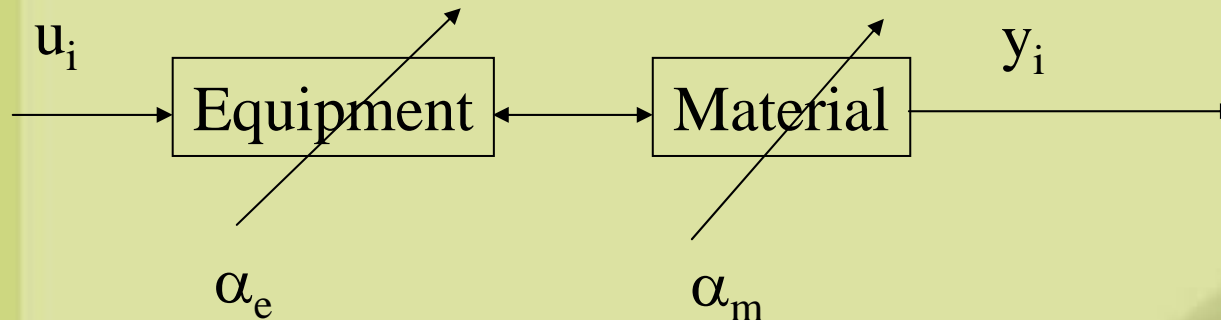


With a Long Sample Time, The Process has no Apparent Dynamics, i.e. a Very Small Time Constant

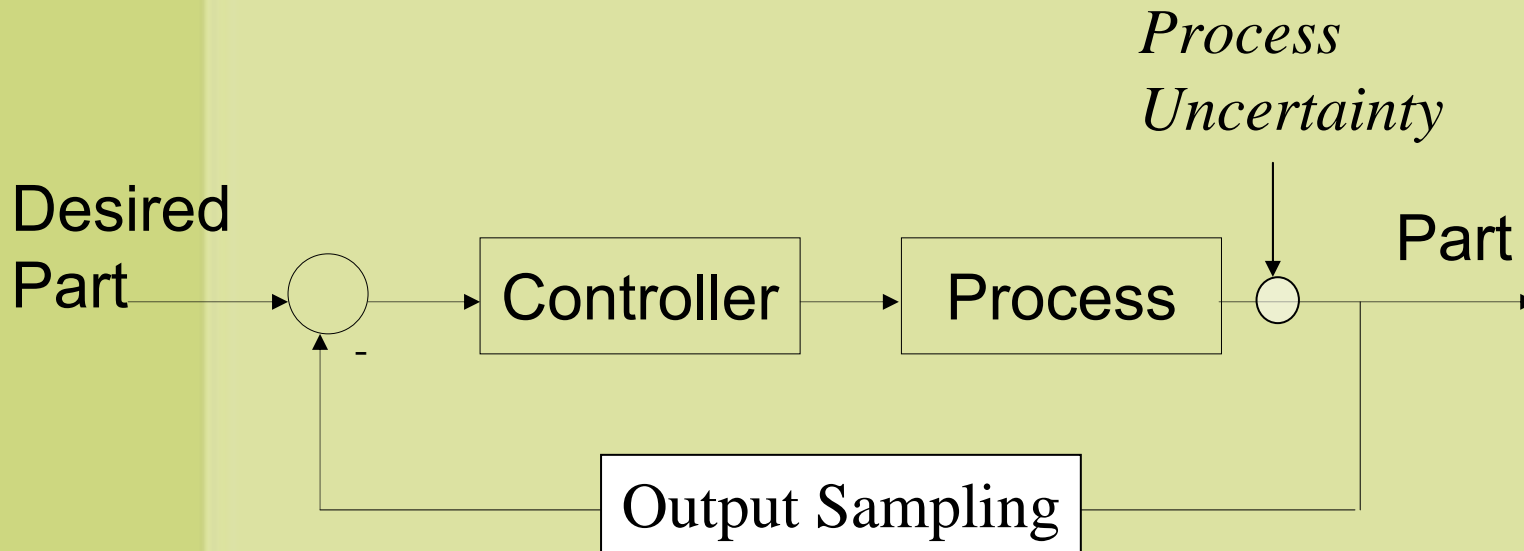
# A Cycle to Cycle Process Model



- or -



# Cycle to Cycle Output Control



# Delays

- **Measurement Delays**
  - Time to acquire and gage
  - Time to reach equilibrium
- **Controller Delays**
  - Time to “decide”
  - Time to compute
- **Process Delay**
  - Waiting for next available machine cycle

# Delays

$z^n = n$  - step time advance operator

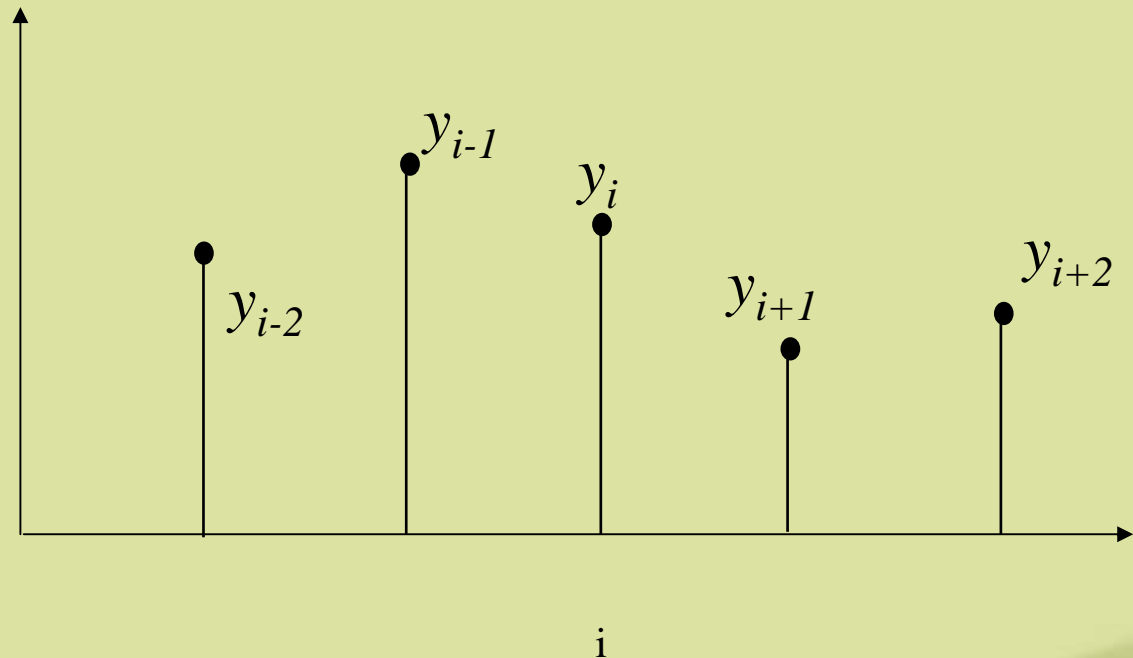
*e.g.*

$$z^1 * y_i = y_{i+1}$$

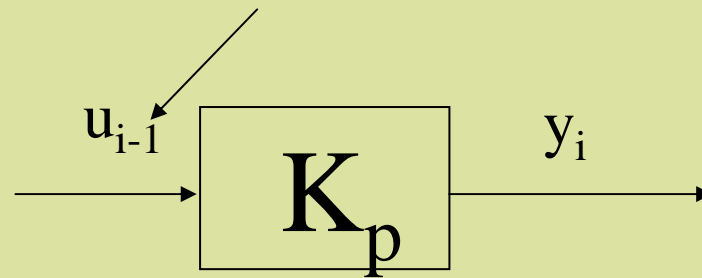
$$z^2 * y_i = y_{i+2}$$

*and*

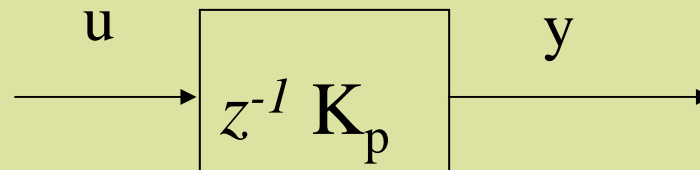
$$z^{-1} * y_i = y_{i-1}$$



# A Pure Delay Process Model

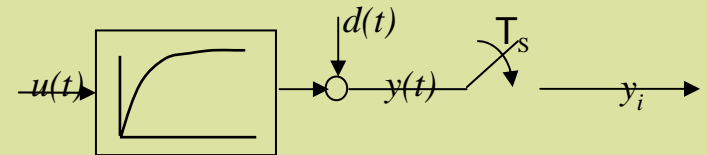
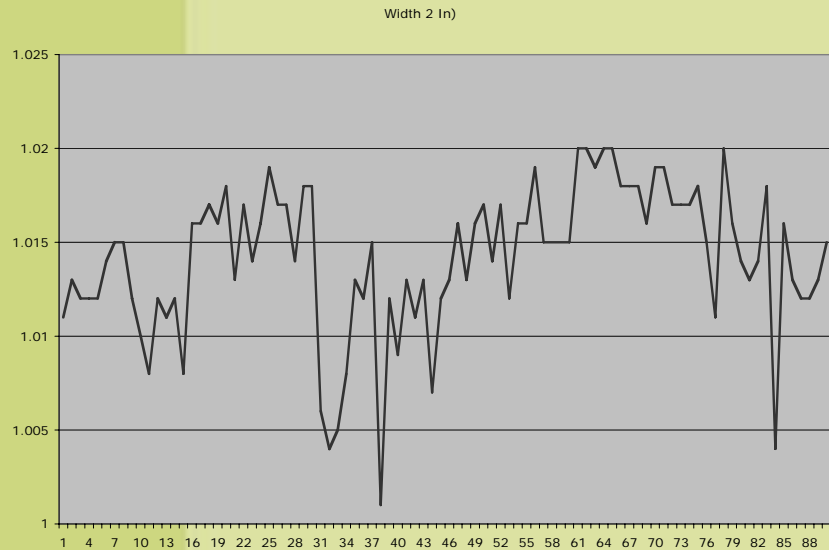


$$y_i = K_p u_{i-1}$$



# Modeling Randomness

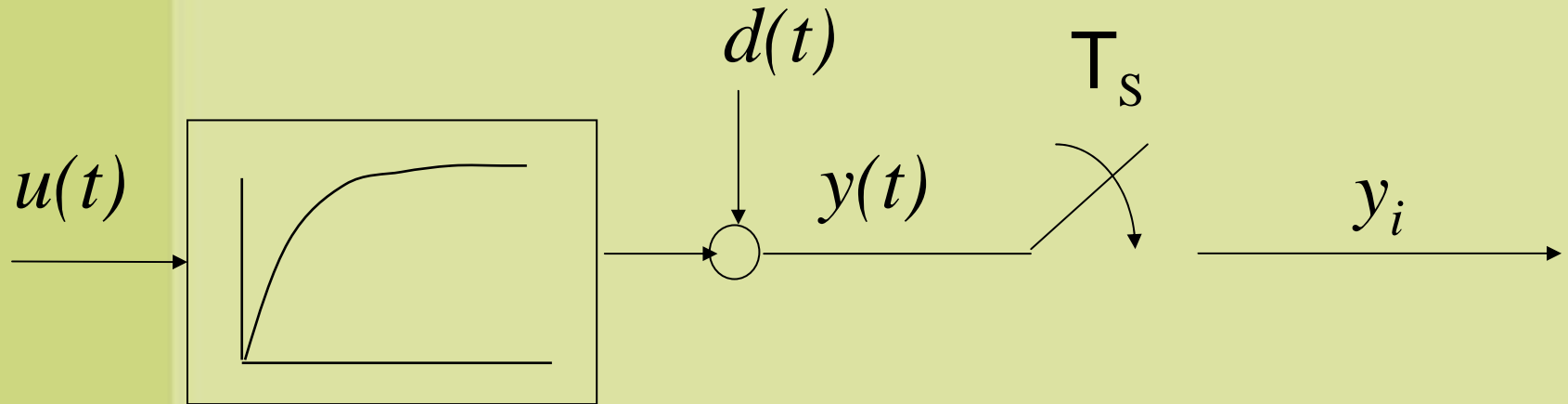
- Recall the Output of a “Real Process”



- Random even with inputs held constant



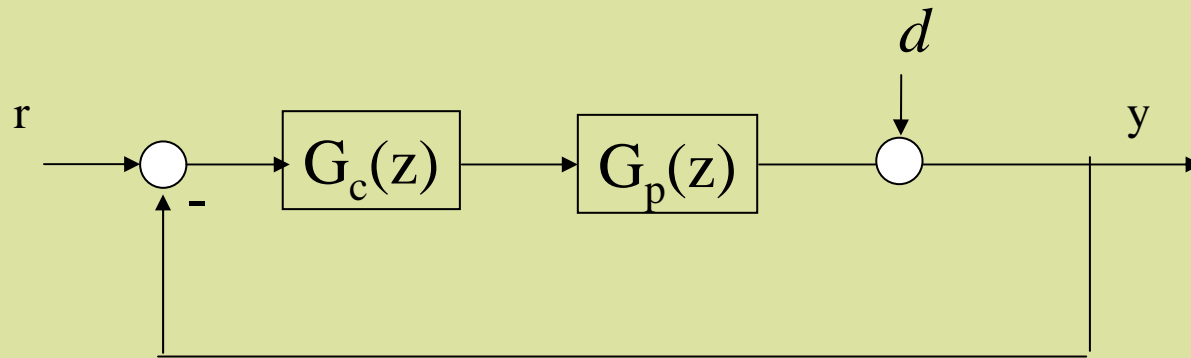
# Output Disturbance Model



Model:

$d(t)$  is a continuous random variable that we sample every cycle ( $T_c$ )

# Or In Cycle to Cycle Control Terms



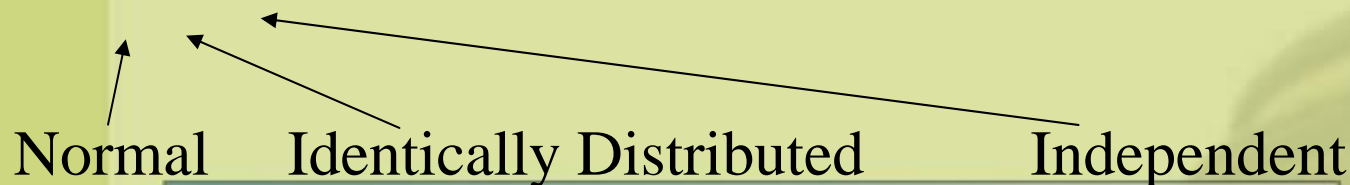
Where:

$d(t)$  is a *sequence of random* numbers governed by a stationary normal distribution function

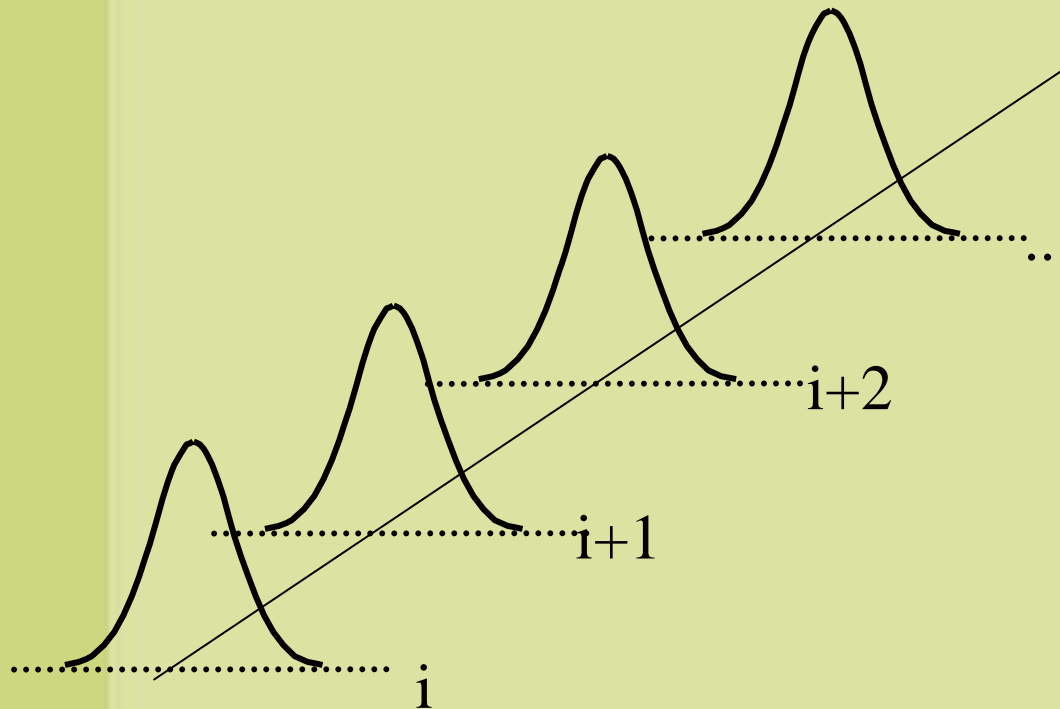
# Gaussian White Noise

- A continuous random variable that at any instant is governed by a normal distribution
- From instant to instant there is no correlation
- Therefore if we sample this process we get:
- A NIDI random number

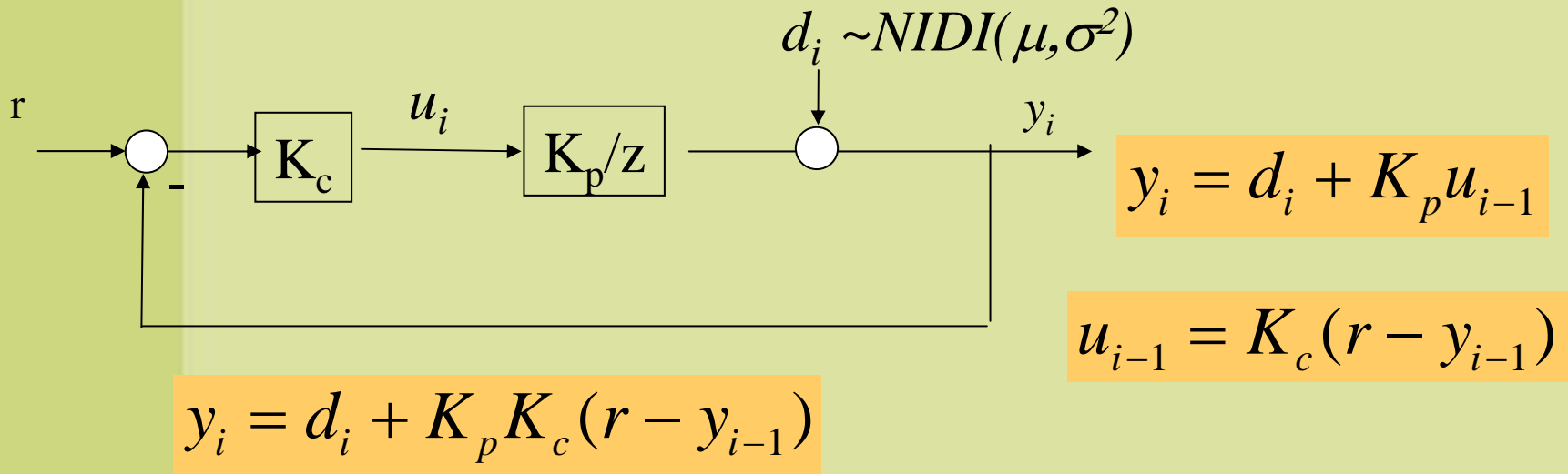
Normal      Identically Distributed      Independent



# The Gaussian “Process”



# Constant (Mean Value) Disturbance Rejection- P control



if  $d_i = \mu$  (a constant), we can look at steady - state behavior:

$$y_i \Rightarrow y_{i-1} \Rightarrow y_\infty = \frac{d_i}{1 + K_p K_c} + r \frac{K_p K_c}{1 + K_p K_c}$$

# And For Example

Thus if we want to eliminate the constant (mean) component of the disturbance

$$\frac{y_{\infty}}{d_i} = \frac{1}{1 + K_p K_c} = \frac{1}{1 + K}$$

Higher loop gain  $K$  improves “rejection”

but only

$K = \infty$  eliminates mean shifts

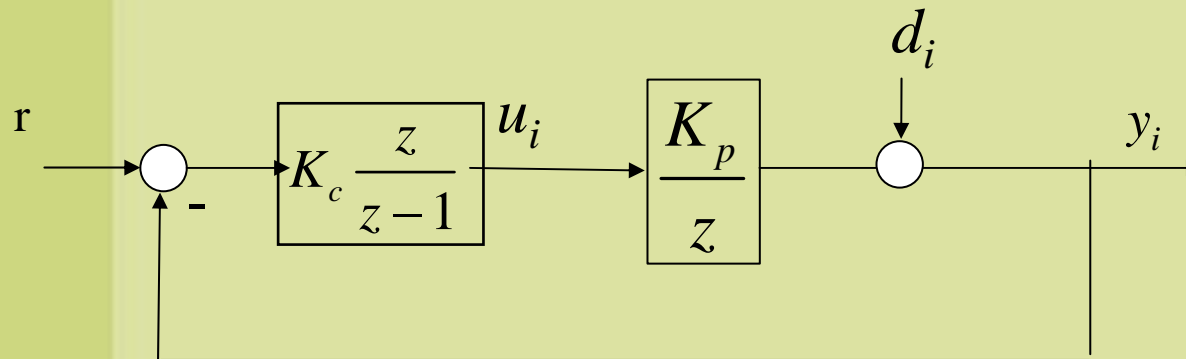
# Error: Try an Integrator

$$u_i = K_c \sum_{j=1}^i e_j \quad \text{running sum of all errors}$$

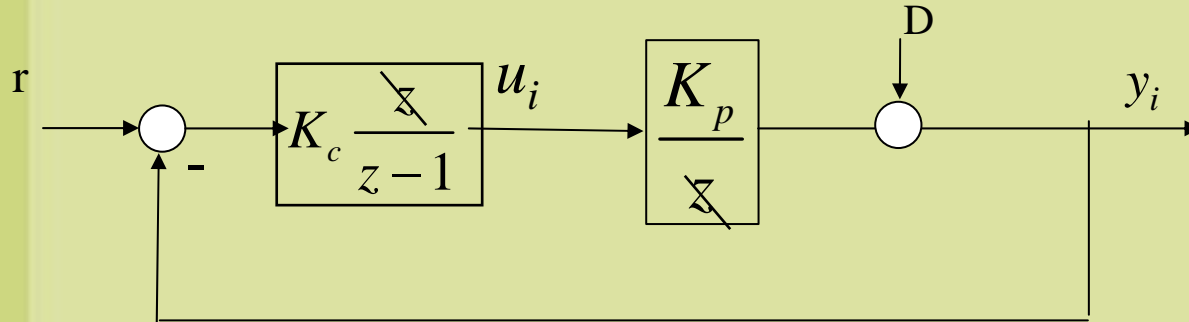
$$u_{i+1} = u_i + K_c e_{i+1} \quad \text{recursive form } (e_i = r - y_i)$$



$$zU = U + K_c zE \quad \Rightarrow \quad G_c(z) = K_c \frac{z}{z-1} = \frac{u}{e}$$



# Constant Disturbance - Integral Control



$$Y(z) = \frac{z-1}{z-1 + K_c K_p} D \quad (\text{Assume } r=0)$$

or  $y_{i+1} + (1 - K_c K_p)y_i = d_{i+1} - d_i$

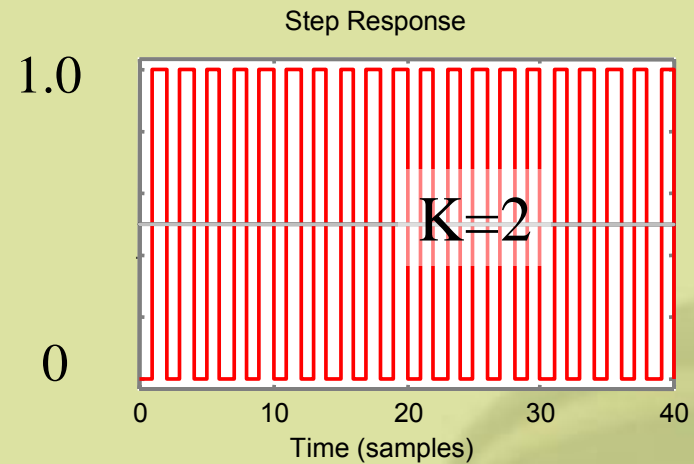
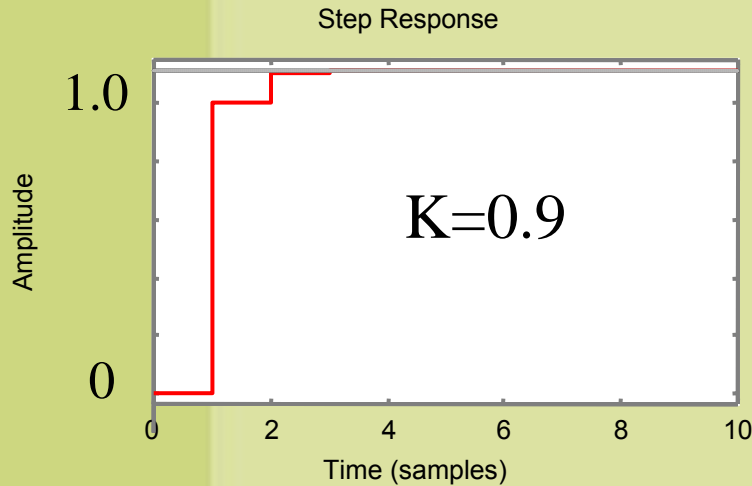
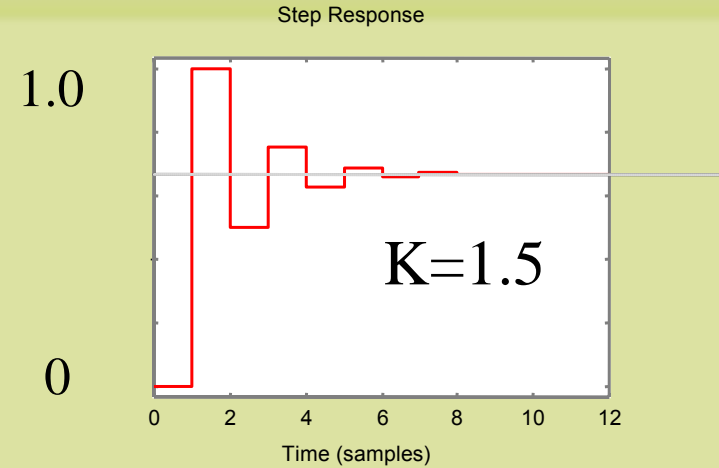
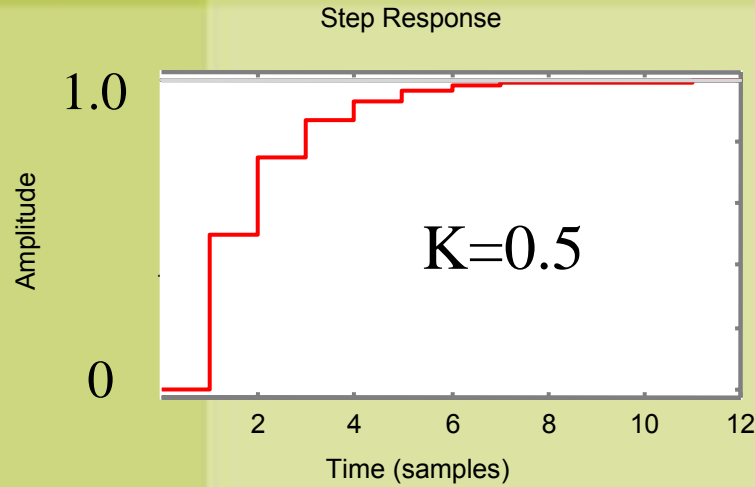
Again at steady state  $y_{i+1} = y_i = y_\infty$

And since  $D$  is a constant  $y_\infty(2 - K_c K_p) = 0$

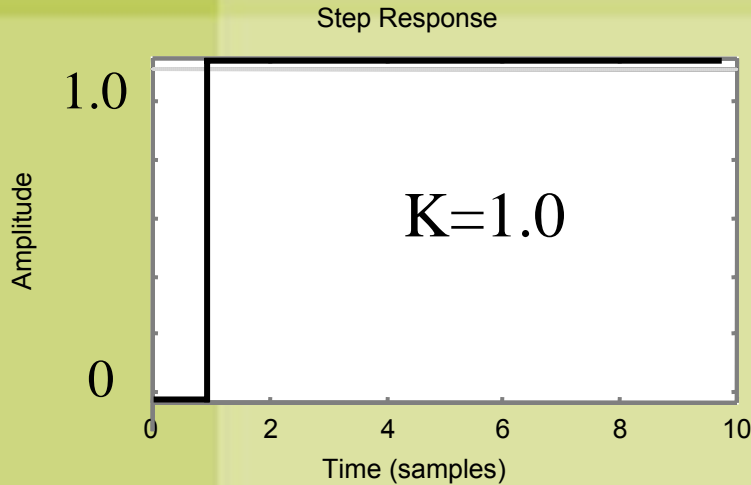
Zero error  
regardless  
of loop  
gain



# Effect of Loop Gain $K$ on Time Response: I-Control



# Effect of Loop Gain $K = K_c K_p$



Best performance at Loop Gain  $K = 1.0$

Stability Limits on Loop Gain  $0 < K < 2$

# What about random component of $d$ ?

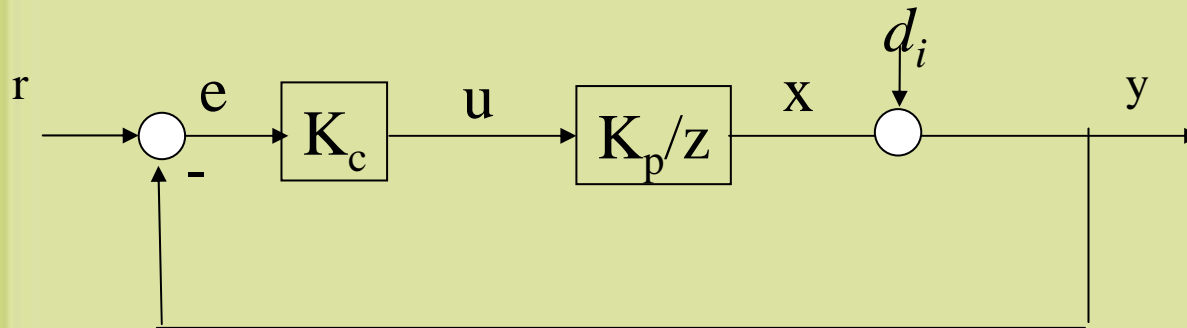
- $d_i$  is defined as a NIDI sequence
- Therefore:
  - Each successive value of the sequence is probably different
  - Knowing the prior values:  $d_{i-1}, d_{i-2}, d_{i-3}, \dots$  will not help in predicting the next value

e.g.  $d_i \neq a_1 d_{i-1} + a_2 d_{i-2} + a_3 d_{i-3} + \dots$



# Thus

This implies that with our cycle to cycle process model under *proportional* control:



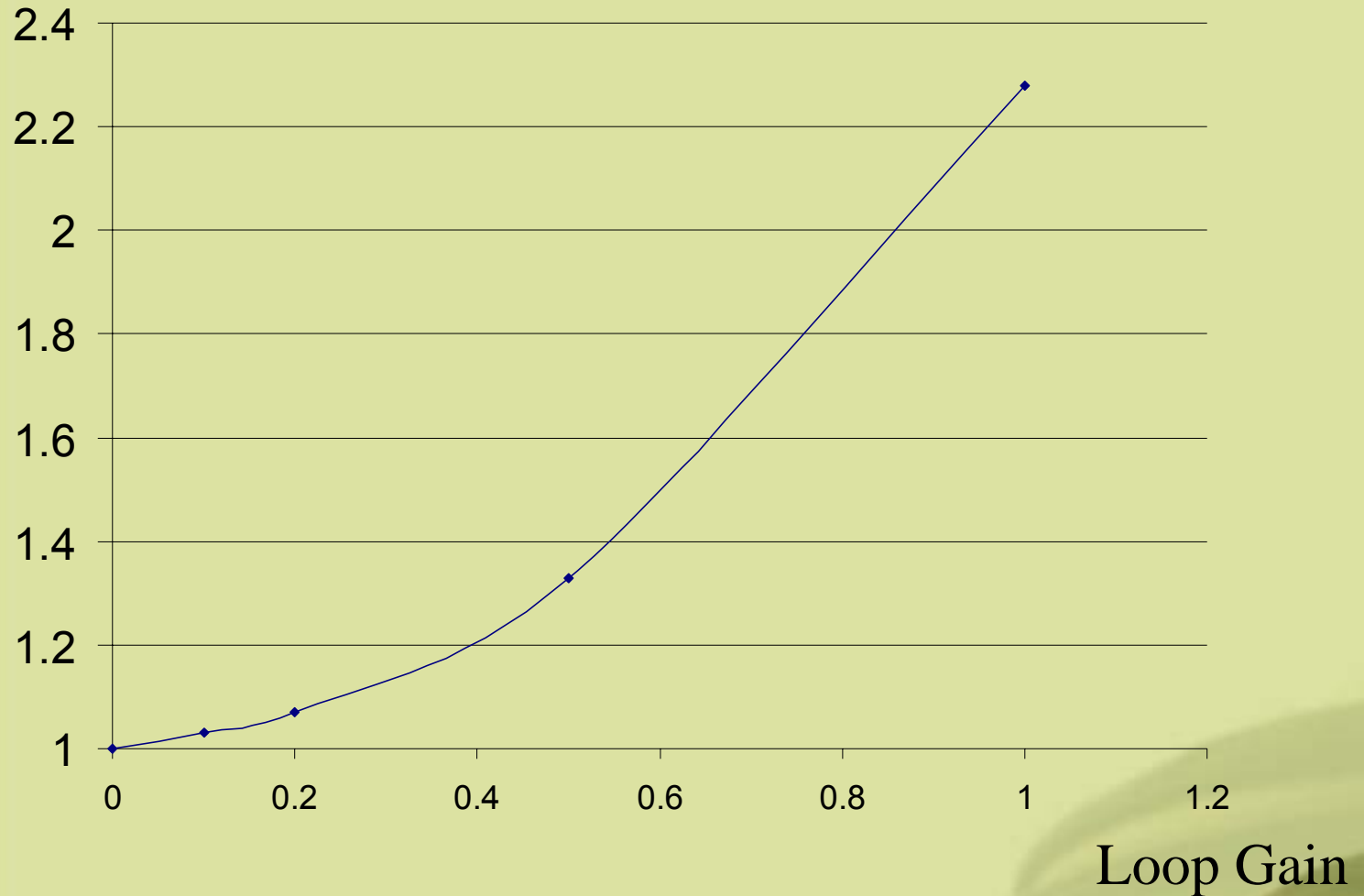
The output of the plant  $x_i$  will *at best* represent the error from the previous value of  $d_{i-1}$

$$x_i = -K_c K_p d_{i-1}$$

will not cancel  $d_i$

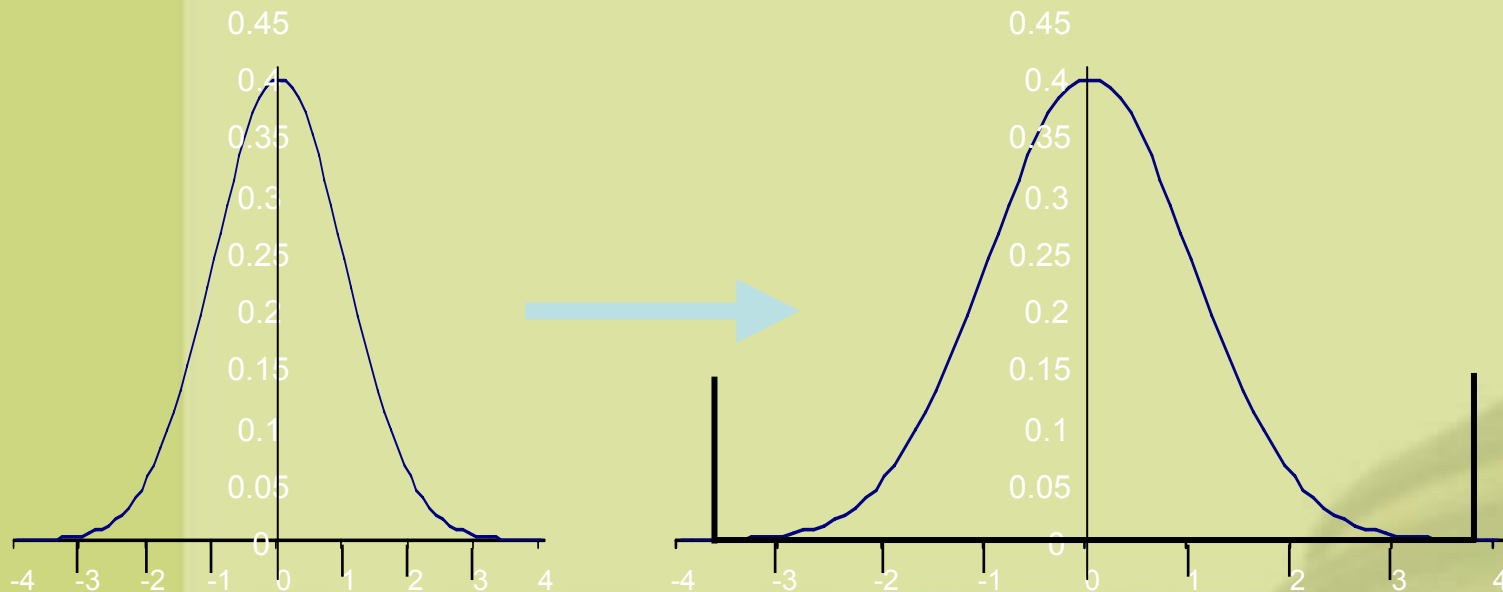
# Variance Change with Loop Gain

$$\frac{\sigma_{CtC}^2}{\sigma^2}$$

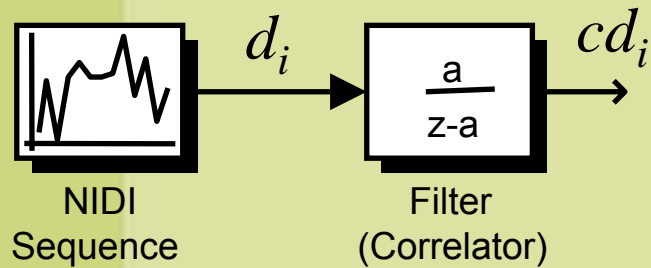


# Conclusion - CtC with Un-Correlated (Independent) Random Disturbance

- Mean error will be zero using “I” control
- Variance will increase with loop gain
- Increase in  $\sigma$  at  $K=1 \sim 1.5 * \sigma_{\text{open loop}}$



# What if the Disturbance is not NIDI?



$$CD(Z)(z - a) = aD(z)$$

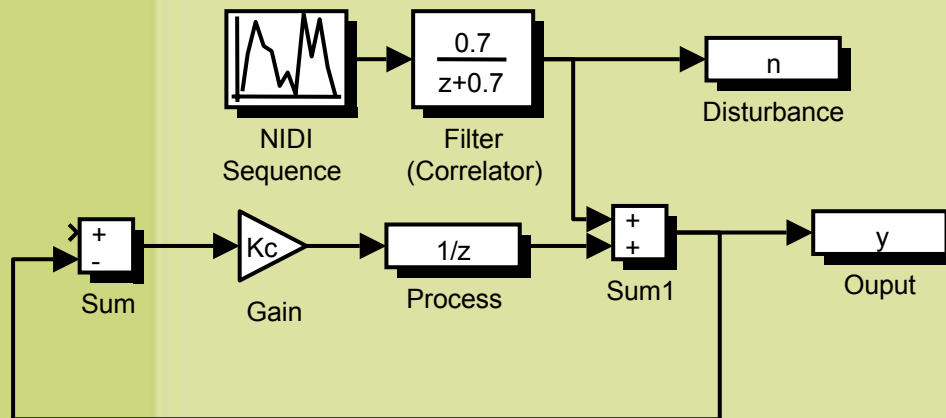
$$cd_{i+1} = a(d_i - cd_i)$$

unknown      known

expect some correlation, therefore ability to counteract some of the disturbances

# What if the Disturbance is not NIDI?

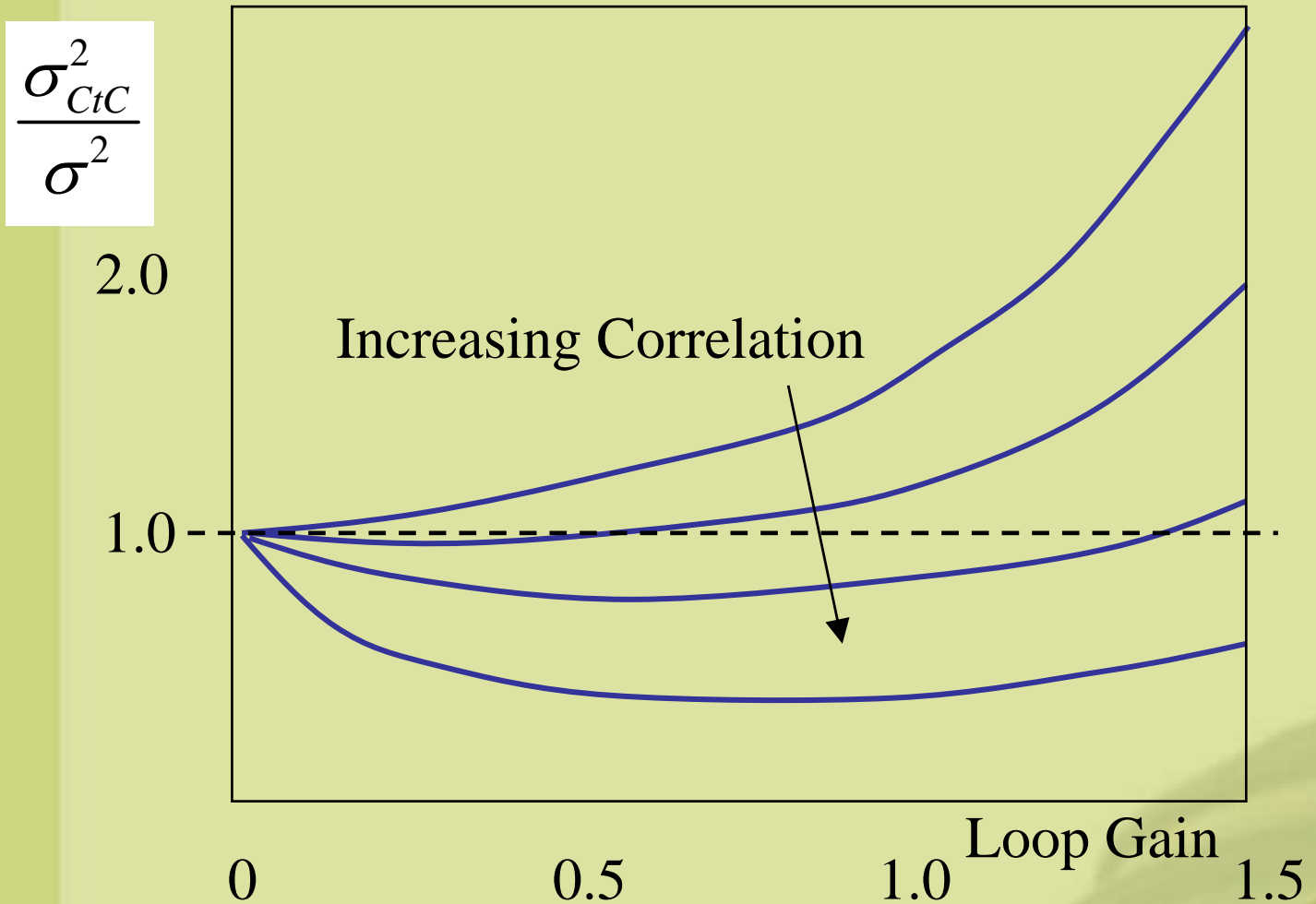
## Proportional Control Simulation



$K_c$	$\sigma^2_{Ctc} / \sigma^2_o$
0	1
0.1	0.89
0.25	0.77
0.5	0.69
0.9	1.39

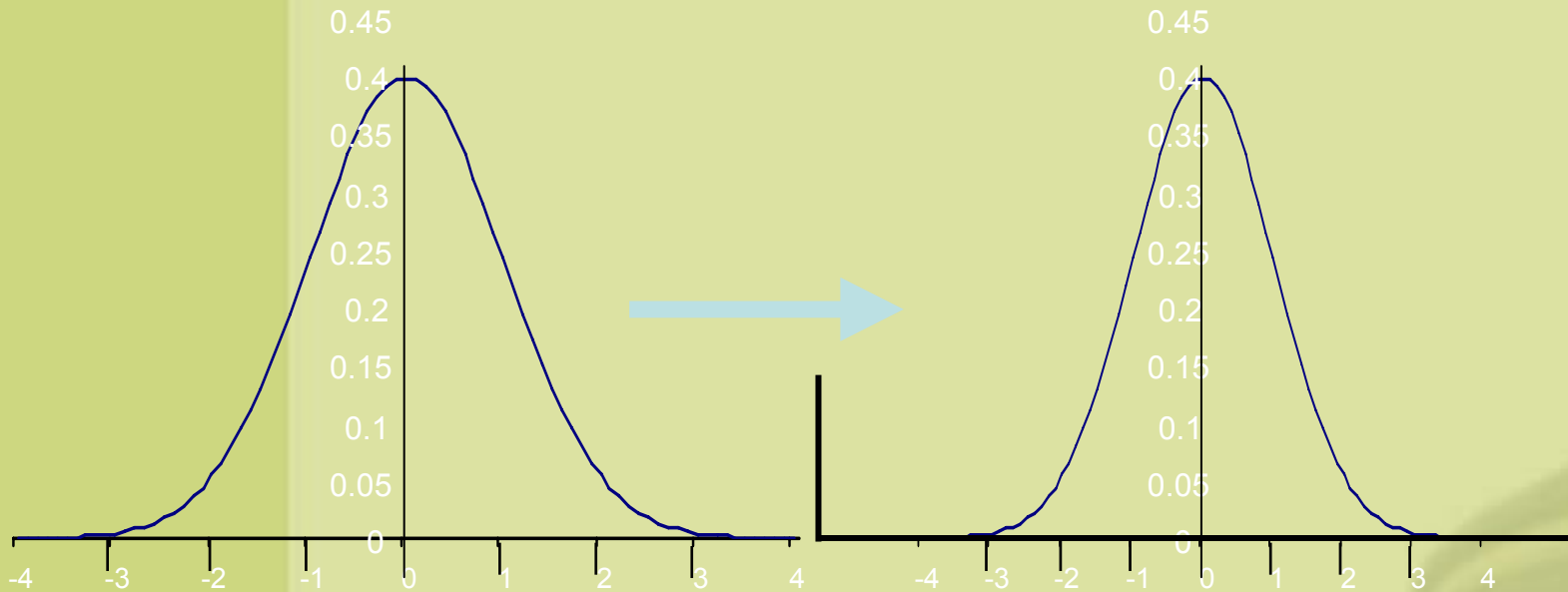


# Gain - Variance Reduction



# Conclusion - CtC with Correlated (Dependent) Random Disturbance

- Mean error will be zero using “I” control
- Variance will decrease with loop gain
- Best Loop Gain is still  $K_c K_p = 1$



# Conclusions from Cycle to Cycle Control Theory

- Feedback Control of NIDI Disturbance will Increase Variance
  - Variance Increases with Gain
- BUT: If Disturbance is *NID* but not *I*; We CAN Decrease Variance
  - Higher Gains -> Lower Variance
  - Design Problem: Low Error and Low Variance

# How to Tell if Disturbance is Independent

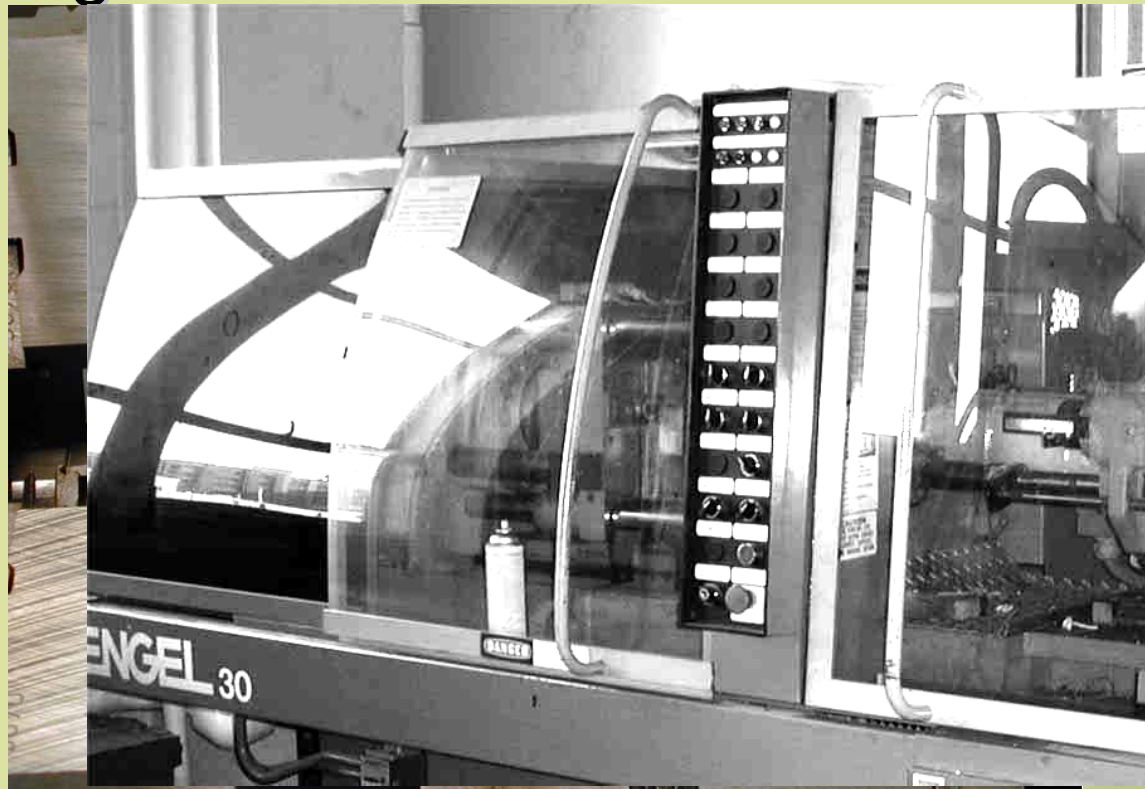
- Correlation of output data
  - Look at the Autocorrelation
  - Effect of Filter on Autocorrelation
- Reaction of Process to Feedback
  - If variance decreases data has dependence

# Is Disturbance is Independent?

- Correlation of output data
  - Look at the Autocorrelation  $\Phi_{xx}(\tau) = \int_{-\infty}^{\infty} x(t)x(t-\tau)dt$
  - Effect of Filter on Autocorrelation
- Reaction of Process to Feedback
  - If variance decreases then data must have some dependence

# But Does It Really Work?

- Let's Look at Bending and Injection Molding



# Experimental Data

**Cycle to Cycle Feedback Control  
of  
Manufacturing Processes**  
by  
George Tsz-Sin Siu

SM Thesis

Massachusetts Institute of Technology

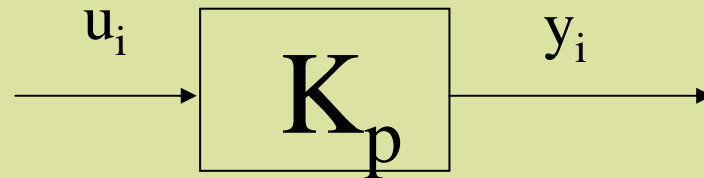
February 2001

# Experimental Results

- Bending
  - Expect NIDI Noise
  - Can Have Step Mean Changes
- Injection Molding
  - Could be Correlated owing to Thermal Effects
  - Step Mean Changes from Cycle Disruption



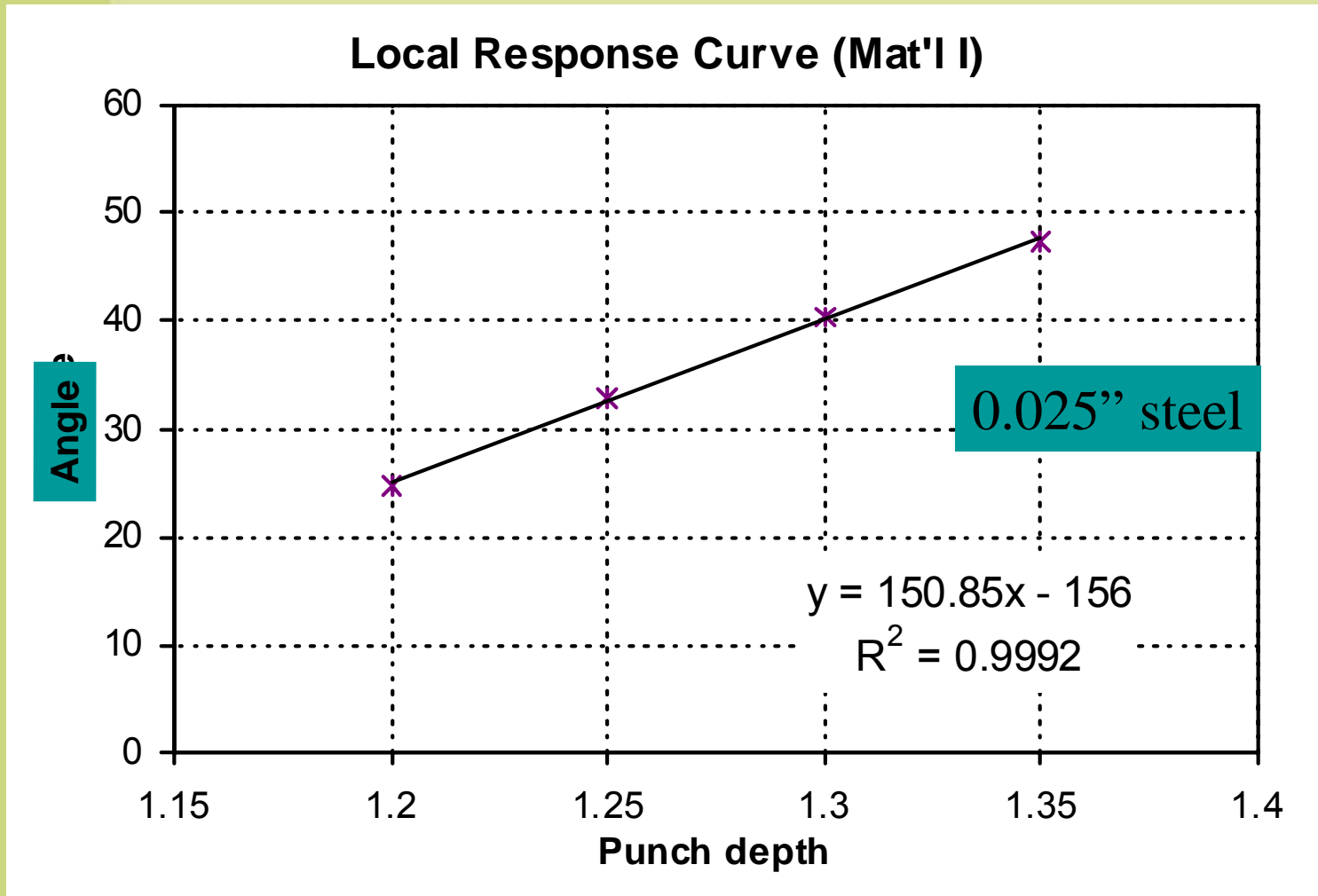
# Process Model for Bending



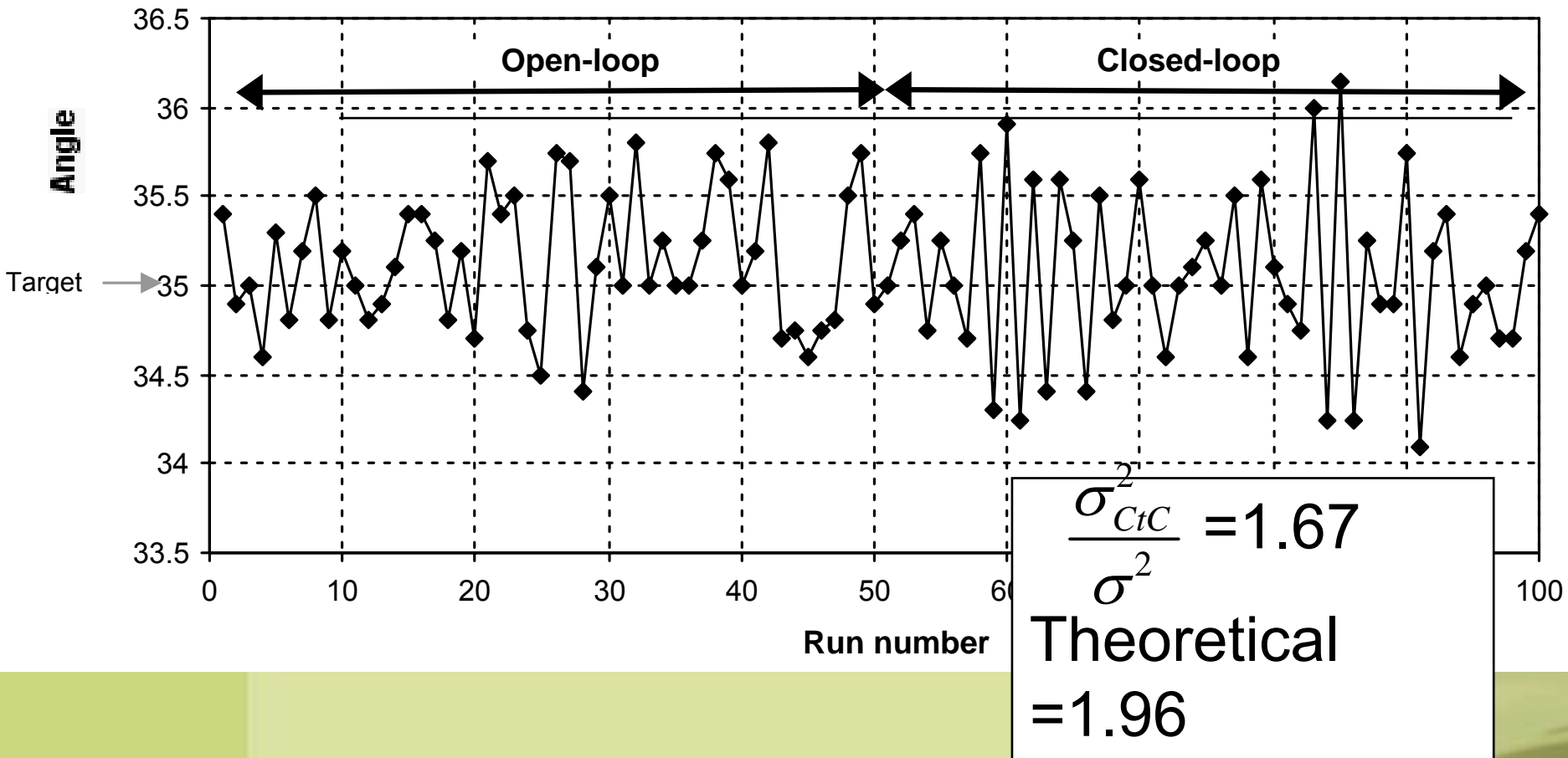
$$y_i = K_p u_{i-1}$$

$$Y(z) = \frac{K_p}{z} \quad K_p = ?$$

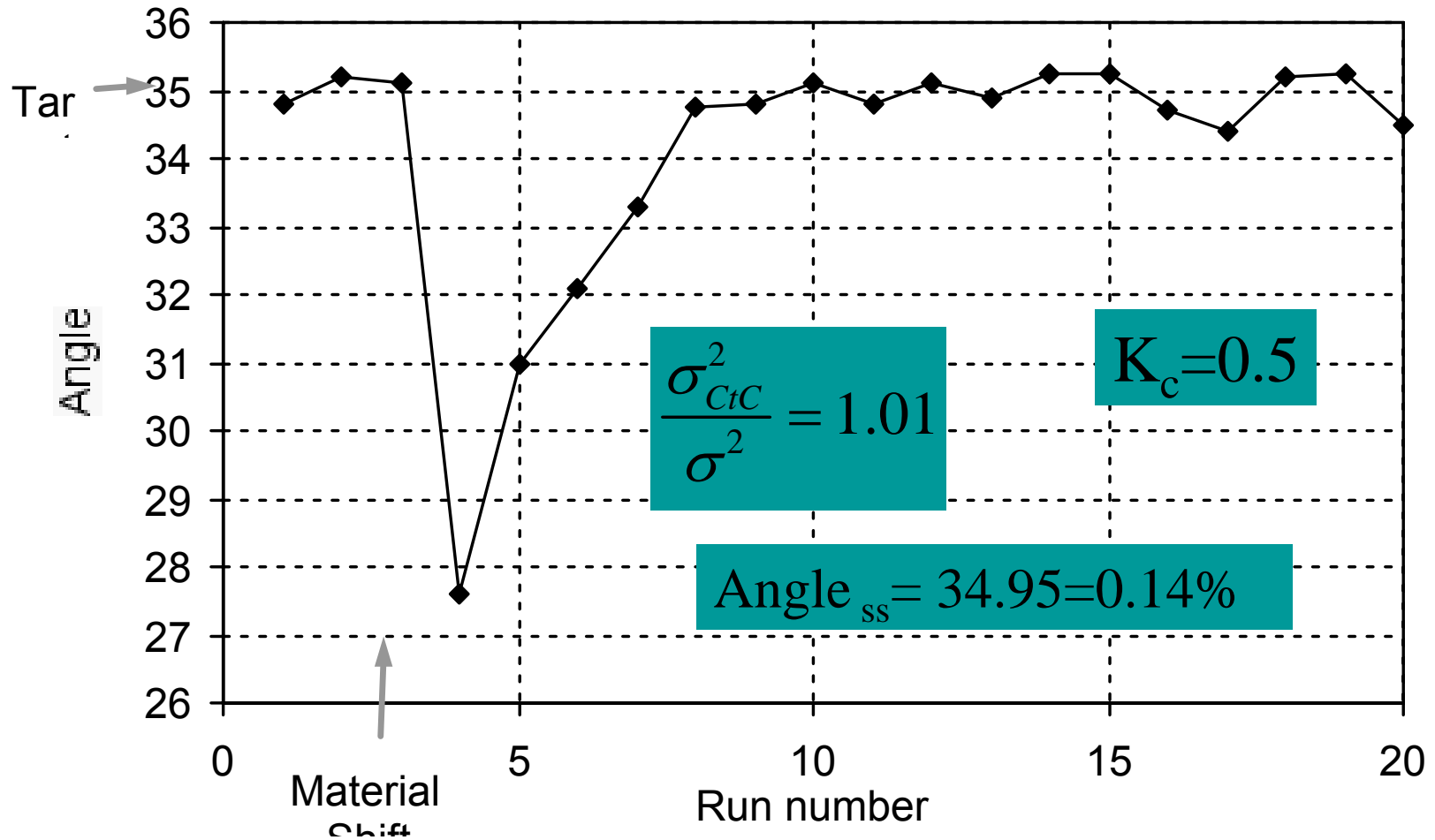
# Process Model for Bending



# Results for $K_C=0.7; \Delta\mu=0$



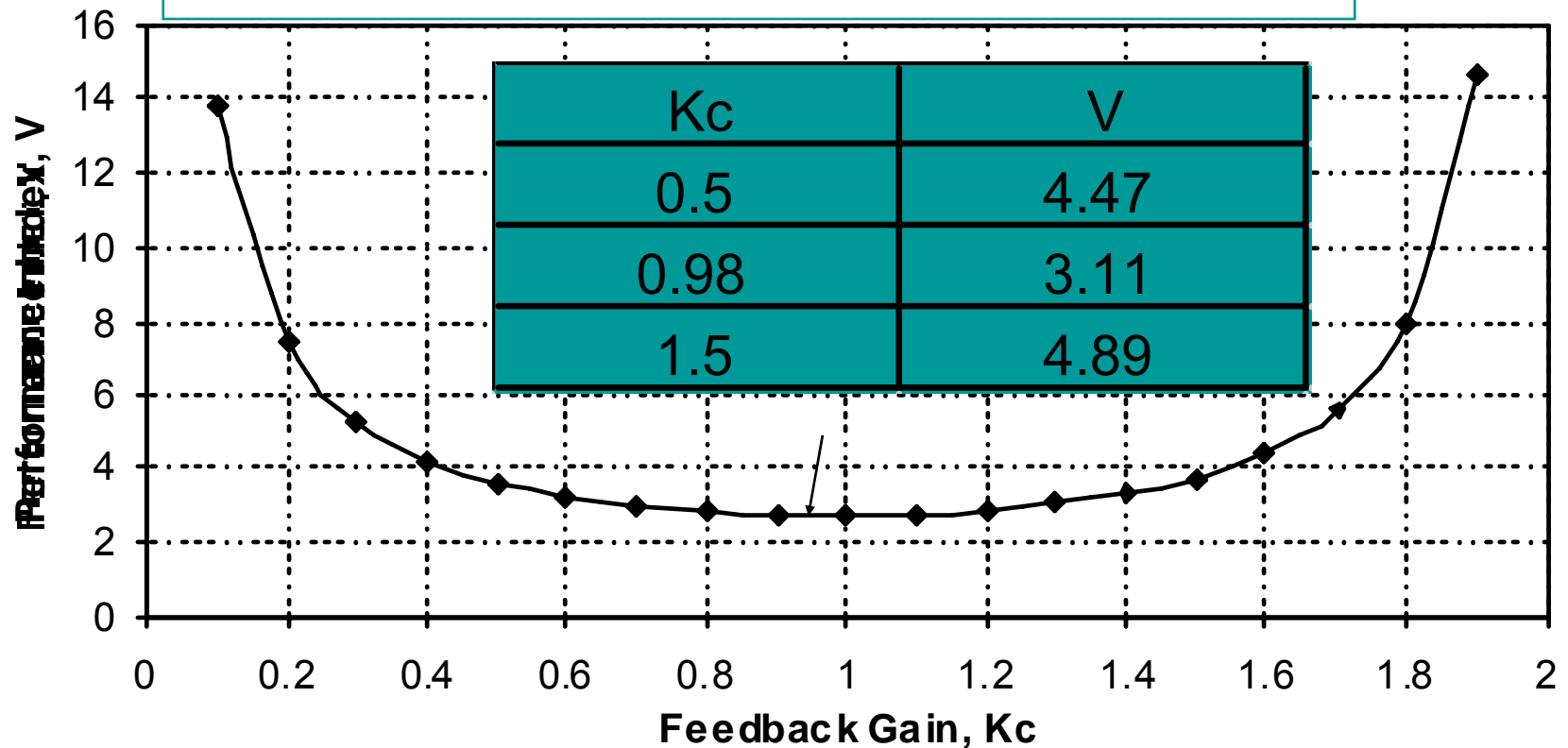
# I-Control $\Delta\mu \neq 0$



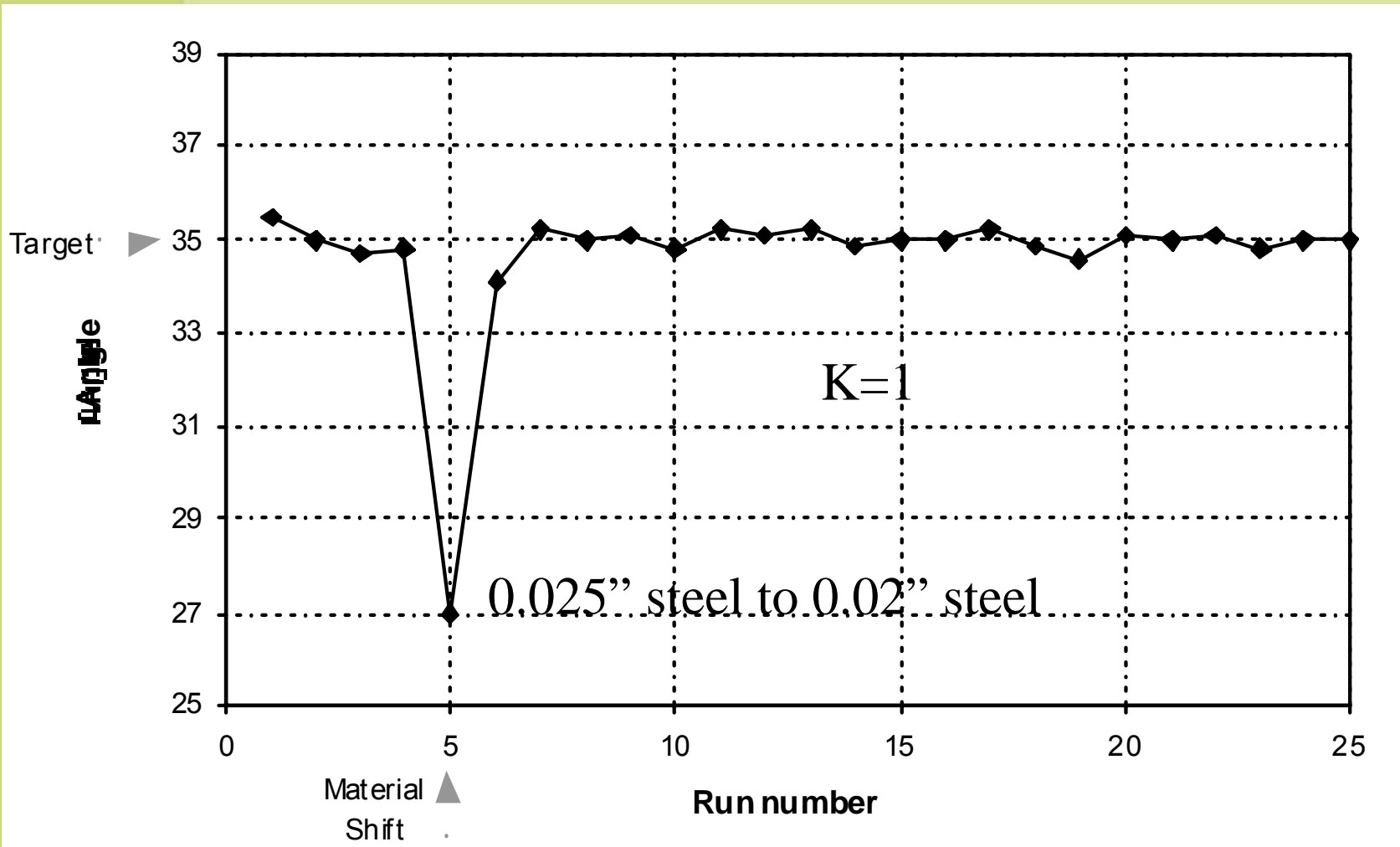
# Minimum Expected Loss Integral-Controller

Calculated Expected Loss vs. Gain

Experimental Verification



# Disturbance Response for "Optimal" Integral Control Gain



# Injection Molding: Process Model

$$\hat{Y} = \beta_0 + \beta_2 \cdot X_2 + \beta_3 \cdot X_3 + \beta_{23} \cdot X_2 \cdot X_3 \quad \text{Initial Model}$$

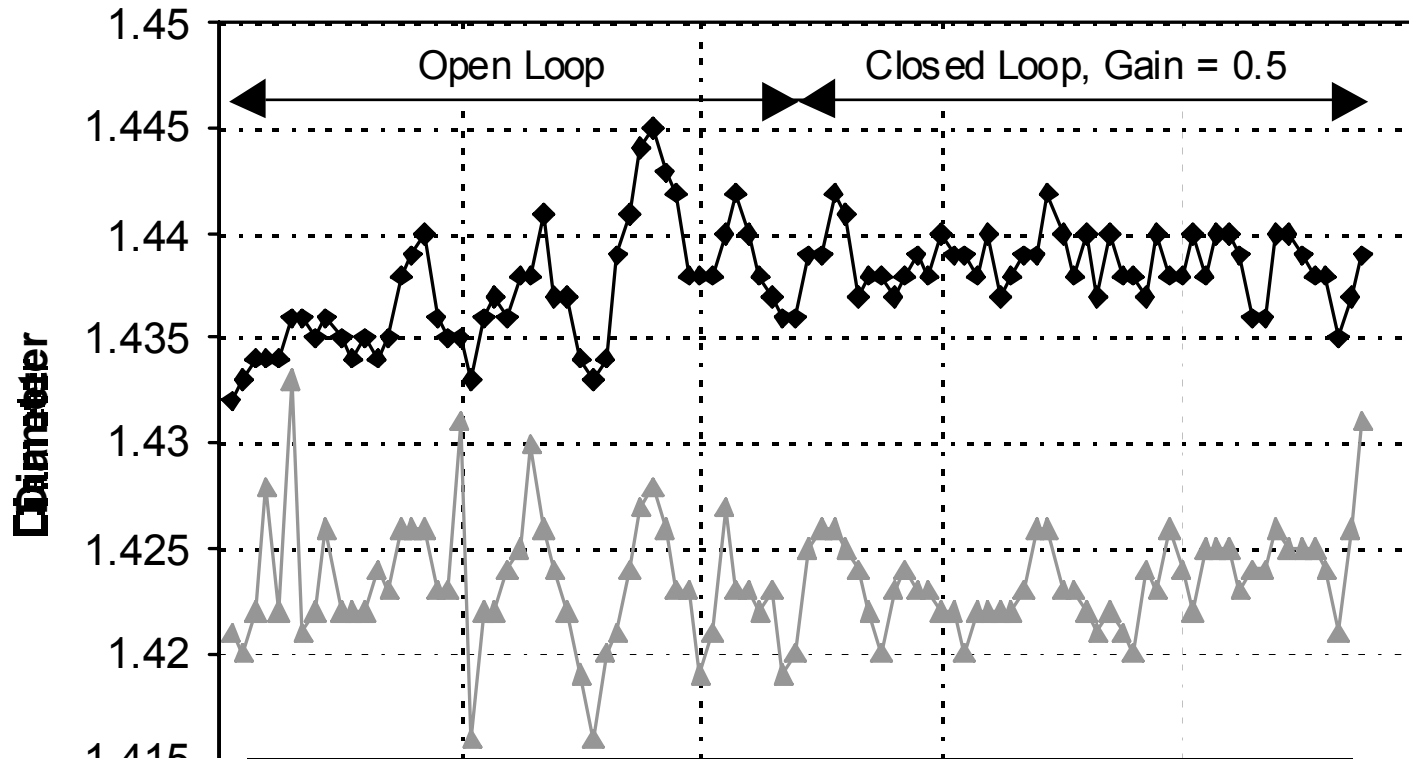
Process inputs	Levels	
X2 = Hold time (seconds)	5 sec	20 sec
X3 = Injection speed (in/sec)	0.5 in/sec	6 in/sec

## ANOVA on model terms

Effect	beta	SS	DOF	MS	F	Fcrit	p-value
1	1.437	49.6	1	49.568	2E+07	4.35	0
X2 (Hold time)	-1.04E-03	0	1	2.60E-05	10.593	4.35	0.004
X3 (Injection speed)	-3.75E-04	0	1	3.38E-06	1.373	4.35	0.255
X2X3 (Hold time*Injection speed)	2.92E-04	0	1	2.04E-06	0.831	4.35	0.373
Error		0	20	2.46E-06			
Total		49.6	24				

$$\hat{Y} = \beta_0 + \beta_2 \cdot X_2 \quad \text{Final Model}$$

# P-Control Injection Molding



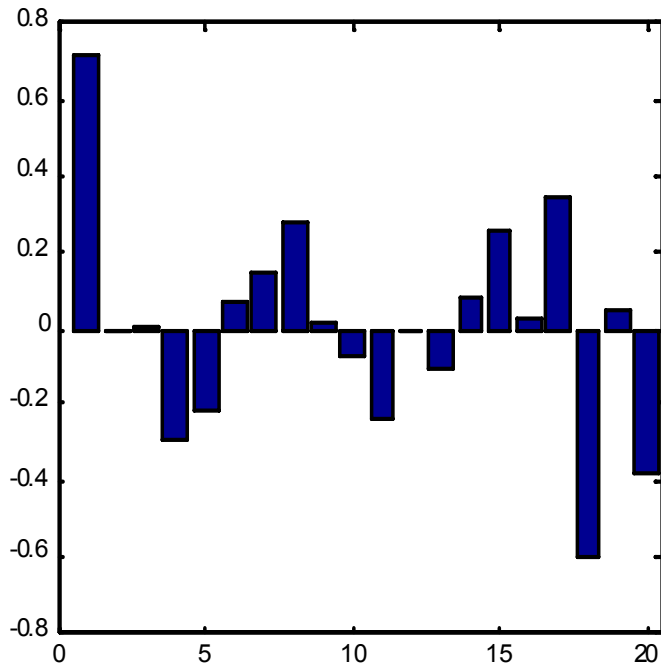
Hot

Cold

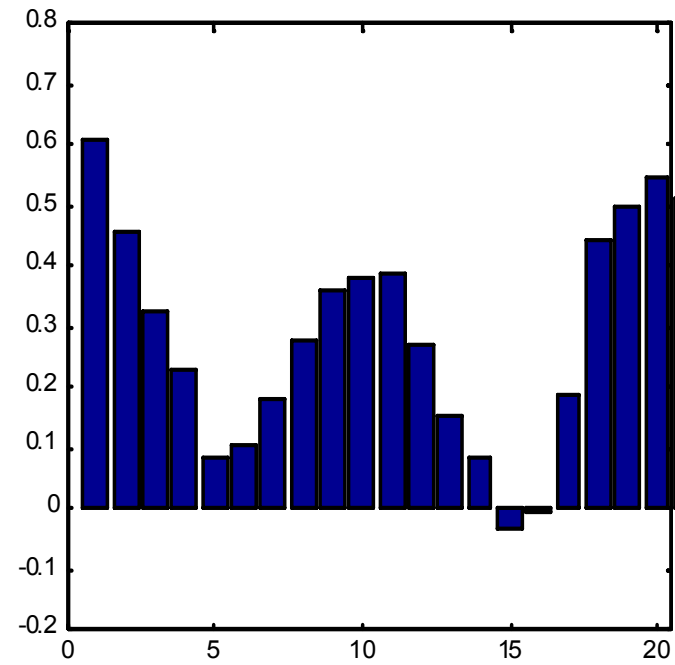
Experiment	Mean(Hot measurement)	Variance	Variance Ratio
Open-loop	1.437	9.97E-06	
Closed-loop	1.439	2.34E-06	0.234



# Output Autocorrelation



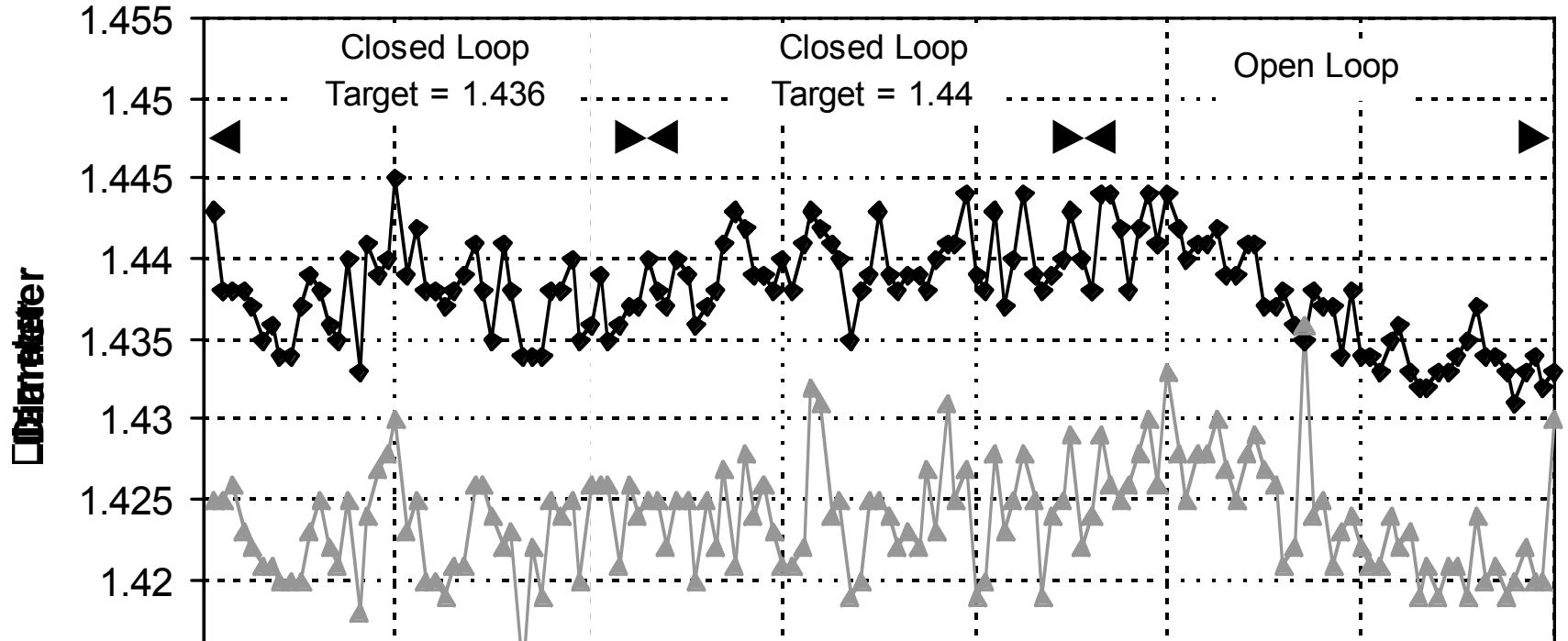
**Bending**



**Injection Molding**

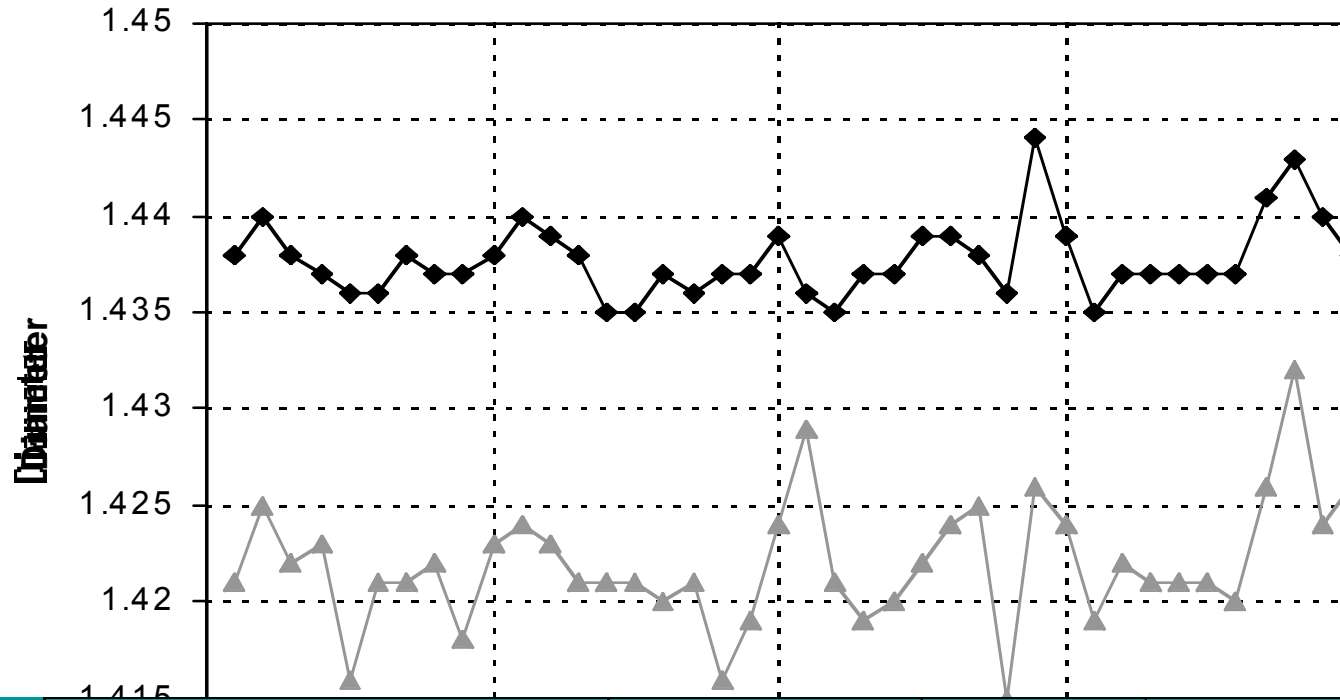
Matlab function XCORR

# P-control: Moving Target



Experiment	Mean (Hot measurement)	Variance	Variance Ratio
Closed-loop, target = 1.436	1.438	6.79E-06	0.471
Closed-loop, target = 1.44	1.440	4.54E-06	0.315
Open-loop	1.437	1.44E-05	-

# Injection Molding: Integral Control



Experiment	Mean (Hot measurement)	Variance	Variance Ratio
Closed-loop, target = 1.436	1.438	3.94E-06	0.395
Open-loop, from first experiment	1.437	9.97E-06	-

# Conclusion

- Model Predictions and Experiment are in Good Agreement
  - Delay - Gain Process Model
  - Normal - Additive Disturbance
  - Effect of Correlated vs. Uncorrelated (NIDI) Disturbances

# Conclusion

- Cycle to Cycle Control
  - Obeys Root Locus Prediction wrt Dynamics
  - Amplifies NIDI Disturbance as Expected
  - Attenuate non-NIDI Disturbance
  - Can Reduce Mean Error (Zero if I-control)
  - Can Reduce “Open Loop” Expected Loss
  - Correlation Sure Helps!!!!
- Can be Extended to Multivariable Case
  - PhD by Adam Rzepniewski (5/5/05)
    - Developed Theory and demonstrated on 100X100 problem (discrete die sheet forming)