6.003: Signals and Systems

Modulation

December 1, 2011

Modulation

Applications of signals and systems in communication systems.

Example: Transmit voice via telephone wires (copper)



Works well: basis of local land-based telephones.

Wireless Communication

In cellular communication systems, signals are transmitted via electromagnetic (E/M) waves.



For efficient transmission and reception, antenna length should be on the order of the wavelength.

Telephone-quality speech contains frequencies from 200 to 3000 Hz.

How long should the antenna be?

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Telephone-quality speech contains frequencies between 200 Hz and 3000 Hz.

How long should the antenna be?

- $1. < 1 \, mm$
- $2.\ \sim cm$
- $3.\ \sim m$
- $4.\ \sim km$
- $5.~>100\,km$

Wavelength is $\lambda = c/f$ so the lowest frequencies (200 Hz) produce the longest wavelengths

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8 \,\mathrm{m/s}}{200 \,\mathrm{Hz}} = 1.5 \times 10^6 \,\mathrm{m} = 1500 \,\mathrm{km} \,.$$

and the highest frequencies (3000 Hz) produce the shortest wavelengths

$$\lambda = \frac{c}{f} = \frac{3 \times 10^8 \,\mathrm{m/s}}{3000 \,\mathrm{Hz}} = 10^5 \,\mathrm{m} = 100 \,\mathrm{km} \,.$$

On the order of hundreds of miles!

For efficient transmission and reception, the antenna length should be on the order of the wavelength.

Telephone-quality speech contains frequencies between 200 Hz and 3000 Hz.

How long should the antenna be? 5

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A wavelength of $10\,\mathrm{cm}$ corresponds to a frequency of

$$f = \frac{c}{\lambda} \sim \frac{3 \times 10^8 \,\mathrm{m/s}}{10 \,\mathrm{cm}} \approx 3 \,\mathrm{GHz}\,. \label{eq:f_stars}$$

Modern cell phones use frequencies near 2 GHz.



Wireless Communication

Speech is not well matched to the wireless medium.

Many applications require the use of signals that are not well matched to the required media.

signal	applications
audio	telephone, radio, phonograph, CD, cell phone, MP3
video	television, cinema, HDTV, DVD
internet	coax, twisted pair, cable TV, DSL, optical fiber, E/M
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We can often modify the signals to obtain a better match.

Today we will introduce simple matching strategies based on **modulation**.





Amplitude Modulation

Multiplying a signal by a sinusoidal **carrier** signal is called amplitude modulation (AM). AM shifts the frequency components of X by $\pm \omega_c$.



Amplitude Modulation

Multiplying a signal by a sinusoidal **carrier** signal is called amplitude modulation. The signal "modulates" the amplitude of the carrier.



Amplitude Modulation



Synchronous Demodulation

X can be recovered by multiplying by the carrier and then low-pass filtering. This process is called **synchronous demodulation**.

$$y(t) = x(t)\cos\omega_c t$$

$$z(t) = y(t)\cos\omega_c t = x(t) \times \cos\omega_c t \times \cos\omega_c t = x(t)\left(\frac{1}{2} + \frac{1}{2}\cos(2\omega_c t)\right)$$

Synchronous Demodulation

Synchronous demodulation: convolution in frequency.



Synchronous Demodulation

We can recover X by low-pass filtering.



Multiple transmitters can co-exist, as long as the frequencies that they transmit do not overlap.



Multiple transmitters simply sum (to first order).



The receiver can select the transmitter of interest by choosing the corresponding demodulation frequency.



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Broadcast Radio

"Broadcast" radio was championed by David Sarnoff, who previously worked at Marconi Wireless Telegraphy Company (point-to-point).

- envisioned "radio music boxes"
- analogous to newspaper, but at speed of light
- receiver must be cheap (as with newsprint)
- transmitter can be expensive (as with printing press)

The problem with making an inexpensive radio receiver is that you must know the carrier signal exactly!



The problem with making an inexpensive radio receiver is that you must know the carrier signal exactly!



What happens if there is a phase shift ϕ between the signal used to modulate and that used to demodulate?

$$y(t) = x(t) \times \cos(\omega_c t) \times \cos(\omega_c t + \phi)$$
$$= x(t) \times \left(\frac{1}{2}\cos\phi + \frac{1}{2}\cos(2\omega_c t + \phi)\right)$$

Passing y(t) through a low pass filter yields $\frac{1}{2}x(t)\cos\phi$.

If $\phi = \pi/2$, the output is zero!

If ϕ changes with time, then the signal "fades."

AM with Carrier

One way to synchronize the sender and receiver is to send the carrier along with the message.



Adding carrier is equivalent to shifting the DC value of x(t).

If we shift the DC value sufficiently, the message is easy to decode: it is just the envelope (minus the DC shift).

If the carrier frequency is much greater than the highest frequency in the message, AM with carrier can be demodulated with a peak detector.



In AM radio, the highest frequency in the message is 5 kHz and the carrier frequency is between 500 kHz and 1500 kHz.

This circuit is simple and inexpensive.

But there is a problem.

AM with carrier requires more power to transmit the carrier than to transmit the message!



Speech sounds have high crest factors (peak value divided by rms value). The DC offset C must be larger than x_p for simple envelope detection to work.

The power needed to transmit the carrier can be $35^2 \approx 1000 \times$ that needed to transmit the message.

Okay for broadcast radio (WBZ: 50 kwatts).

Not for point-to-point (cell phone batteries wouldn't last long!).

Envelope detection also cannot separate multiple senders.





Superheterodyne Receiver

Edwin Howard Armstrong invented the superheterodyne receiver, which made broadcast AM practical.



Edwin Howard Armstrong also invented and patented the "regenerative" (positive feedback) circuit for amplifying radio signals (while he was a junior at Columbia University). He also invented wide-band FM.

Digital Radio

Could we implement a radio with digital electronics?

Commercial AM radio

- 106 channels
- each channel is allocated 10 kHz bandwidth
- center frequencies from 540 to 1600 kHz







Digital Radio

The digital electronics must implement a bandpass filter, multiplication by $\cos \omega_c t$, and a lowpass filter.





$$h[n] = h_{LPF}[n] \cos \Omega_c n$$

$$y[n] = x[n] * (h_{LPF}[n] \cos \Omega_c n)$$

$$= \sum_k x[k]h_{LPF}[n-k] \cos \Omega_c (n-k)$$

$$= \sum_k x[k]h_{LPF}[n-k] (\cos \Omega_c n \cos \Omega_c k + \sin \Omega_c n \sin \Omega_c k)$$

$$= \left(\sum_k x[k] \cos \Omega_c k h_{LPF}[n-k]\right) \cos \Omega_c n$$

$$+ \left(\sum_k x[k] \sin \Omega_c k h_{LPF}[n-k]\right) \sin \Omega_c n$$

$$= \left((x[n] \cos \Omega_c n) * h_{LPF}[n]\right) \cos \Omega_c n$$

$$+ \left((x[n] \sin \Omega_c n) * h_{LPF}[n]\right) \sin \Omega_c n$$



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