

Massachusetts Institute of Technology

Department of Electrical Engineering and Computer Science
6.245: MULTIVARIABLE CONTROL SYSTEMS

by A. Megretski

Problem Set 10 (due May 12, 2004) ¹

Problem 10.1

For a cone $\Delta = \{\Delta\}$ of complex n -by- m matrices, and for a complex m -by- n matrix M , the quantity $\mu_{\Delta}(M)$ is defined by

$$\mu_{\Delta}(M) = (\inf\{\|\Delta\| : \Delta \in \Delta, \det(I - M\Delta) = 0\})^{-1}$$

(in particular, $\mu_{\Delta}(M) = 0$ if $I - M\Delta$ is invertible for all $\Delta \in \Delta$). Such quantity, called *structured singular value* of M (where Δ is what defines the “structure”), plays an important role in analysing robust stability.

When Δ is the cone of *all* matrices, $\mu_{\Delta}(M)$ equals the usual largest singular number of M . When Δ is the set of all diagonal matrices with complex entries, $\mu_{\Delta}(M) = \mu_{\mathbf{C}}(M)$ is called the *complex structured singular value*. When Δ is the set of all diagonal matrices with real entries, $\mu_{\Delta}(M) = \mu_{\mathbf{R}}(M)$ is called the *real structured singular value*.

Let Δ be the cone of diagonal matrices with complex entries z_i such that $\operatorname{Re}(z_i) \geq |\operatorname{Im}(z_i)|$. Our objective is to produce a method for estimating $\mu_{\Delta}(M)$, based on semidefinite programming.

- (a) Describe the set of all quadratic constraints which are satisfied for the relation between two complex numbers w and v satisfying $w = zv$, where $\operatorname{Re}(z) \geq |\operatorname{Im}(z)|$.
- (b) Use the result of (a) to develop an LMI optimization algorithm for calculating an upper bound $\hat{\mu}_{\Delta}(M)$ of $\mu_{\Delta}(M)$ for an arbitrary n -by- n complex matrix M .

¹Version of May 6, 2004

- (c) Test the upper bound on a set of randomly generated 3-by-3 and 10-by-10 complex matrices M . Compare $\hat{\mu}_{\Delta}(M)$ with $\mu_{\mathbf{C}}(M)$, which can be estimated using MATLAB's

```
bounds=mu(M)
```

(the two components of output `bounds` will be an upper and a lower bound of $\mu_{\mathbf{C}}(M)$).

Problem 10.2

In the design setup shown on Figure 10.1, r is the reference signal, y is measured plant output, u is control action, and $e = y - r$ is tracking error. Transfer functions W_1 (reference

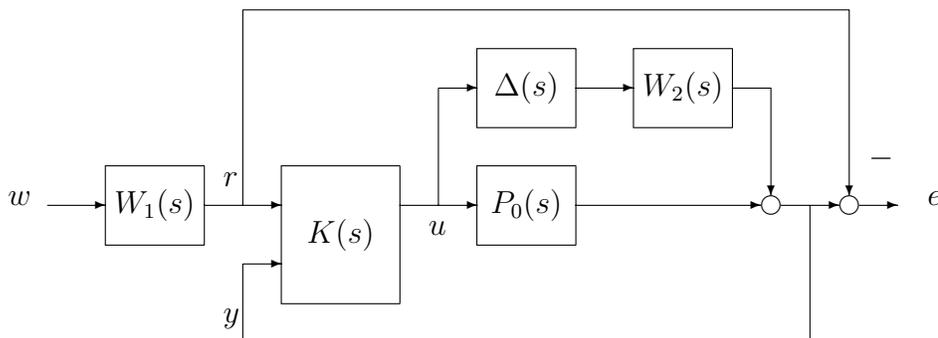


Figure 10.1: Design setup for Problem 10.2

signal shaping filter), P_0 (nominal plant model), and W_2 (uncertainty weight) are given:

$$W_1(s) = \frac{1}{1 + 20s}, \quad W_2(s) = r \frac{s + 1}{s + 10}, \quad P_0(s) = \frac{s - 2}{s^2 - 1},$$

where $r > 0$ is a parameter. $\Delta = \Delta(s)$ is the normalized uncertainty, ranging over the set of all stable transfer functions with $\|\Delta\|_{\infty} \leq 1$. The objective is to design an LTI controller $K = K(s)$ of order not larger than 8, which stabilizes the feedback system for all possible Δ (“robust stabilization”), while trying to make the worst (again, over all possible Δ) closed loop H-Infinity norm from w to e (“robust performance”) as small as possible.

- (a) Find the maximal value r_0 of those $r > 0$ for which robust stabilization is possible.

- (b) For $r = 0.1r_0$, use D-K iterations of H-Infinity optimization and semidefinite programming to minimize robust performance.