6.622 Power Electronics

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# Lecture 5 — Intro to $\mathrm{DC}/\mathrm{DC}$

## 1 First Averaged Circuit Rules

KCL

$$\sum_{j} i_{j} = 0$$

$$\frac{1}{T} \int_{T} \sum_{j} i_{j} dt$$

$$\sum_{j} \frac{1}{T} \int_{T} i_{j} dt$$

$$\sum_{j} \langle i_{j} \rangle = 0$$

KCL applies to time averaged currents (constant charge). The same is true for KVL ( $\sum_k \langle V_k \rangle = 0$ ).

So for a power converter in periodic steady state:

1. Averaged KCL  $\sum_j < i_j >= 0$ 

2. Averaged KVL  $\sum_k < V_k >= 0$ 

3. Capacitor in P.S.S.  $<\lambda_c>=0$ 

4. Inductor in P.S.S.  $\langle V_L \rangle = 0$ 

5. If system lossless (Conservation of energy)  $< P_{in} > = < P_{out} >$ 

### 2 Review



Text on bottom: Switch X on when  $q_x(t) = 1$ , off when  $q_x(t) = 0$ Assuming Ls and Cs are very big:  $v_0(t) = V_0$ ,  $i_2(t) = I_2$ Using average relation in P.S.S.  $\langle v_L \rangle = \langle v_x \rangle - \langle v_2 \rangle$ 

$$\langle v_L \rangle = \frac{1}{T} [DT(v_1 - v_2) + (1 - D)T(-v_2)] = 0 \Rightarrow V_2 = Dv_1$$

#### 2.1 Aside

What do switch 1 and switch 2 do? Let's ignore inductor ripple.



Average power into switch 1:

 $< P_1 > = < (V_Y + v_{Y,AC})(I_Y + i_{Y,AC}) > = < V_Y I_Y > + < v_{Y,AC} i_{Y,AC} >$ 

Where the right side has no cross terms (orthogonal)

$$\langle P_1 \rangle = D(1-D)I_LV_1 + \{D[-(1-D)^2I_LV_1] + (1-D)[-D^2I_LV_1]\}$$

 $= D(1-D)I_LV_1 - D(1-D)I_LV_1 = 0$ 

Where the first term is average power into switch due to  $i_L v$  and the second term is average power into switch due to  $i_{ac,l}$ , v (Could not exactly read this part from original notes)

Switch  $S_1$  takes average power in from the current, voltage and puts equal power out at ac current, voltage. Converts power (efficiently) from dc to ac waveforms! ("inverting" "switch")  $S_2$  does the opposite (converts power from ac waveforms to dc waveforms!) (rectifying switch)

#### 3 Review

Consider input current: L's, C's big,  $i_l(t) \approx I_L$ P.S.S.  $\langle i_{c2} \rangle = 0 \therefore I_2 = I_L$ P.S.S.

$$< I_{c1} >= 0$$
 :  $I_1 =< i_1 >=< i_y >= DI_2$ 

$$I_1 = DI_2$$



Combining with previous result that  $DV_1 = V_2$ :  $I_1V_1 = I_2V_2 \leftarrow$  Lossless system! Note: The trick is to be careful about when one is dealing with <u>instantaneous</u> variables and when one is dealing with <u>average</u> variables!

e.g. at a given instant,  $i_y(t) \neq \langle i_y(t) \rangle$ 

Switch implementation for this case,  $v_1, v_2 > 0$ 



Power flows form 1 to 2.

"Down" or "buck" converter. A type of "direct" converter because in one switch state, power flows directly from input to output.

Suppose we switch source and resistor.



\*note: redefine q(t) = 1 as switch "down" position.

If C's, L's big, same analysis:



$$\langle v_L \rangle = 0 ::\langle v_x \rangle = (1 - D)V_2 = V_1$$
  
 $\therefore V_2 = \frac{V_1}{1 - D} \text{ and } \frac{I_2}{1 - D} = I_1$ 

If  $v_1, v_2 > 0$ , then power flows from  $2 \leftarrow 1$  and  $v_2 > v_1$ 

Boost converter (or "up" converter):



Sometimes drawn  $L \to R$  power flow (But nothing fundamental about it).



 $\ast$  Show boost converter demo circuit built by Katie R. and Sauparna Das



Show:

- 1.  $v_{DS}$
- 2.  $i_L$
- 3.  $v_0$  on scope

Explain boost operation:

- Switch turns on,  $i_L$  rises and incrementally stores energy in L from  $V_I$
- Switch turns off and this energy plus additional energy from  $V_I$  is transformed to output
- Steady state voltages are determined by  $V_2 = \frac{V_1}{1-D}$

Either back or boost can be viewed as a connection of a "canonical cell"



- Direct connection has B common
- one <u>cannot</u> tell power flow direction without knowing
  - 1. external networks
  - 2. switch implementation
  - 3. control

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