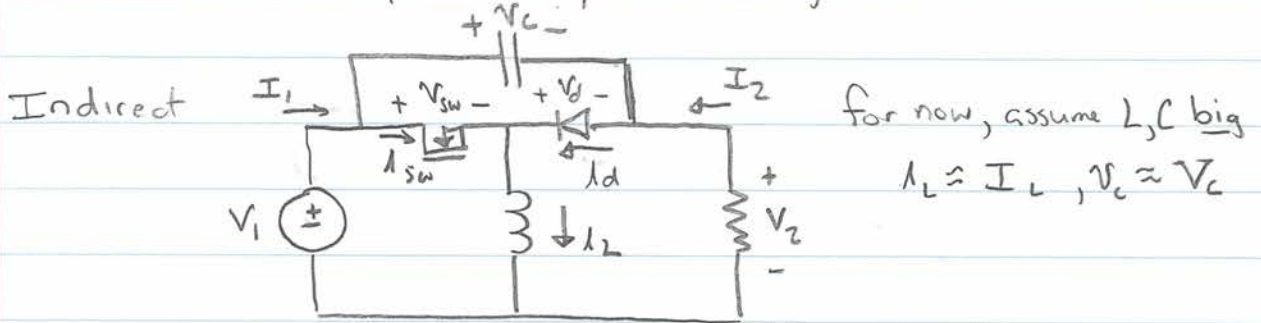


Consider Device and passive component ratings of converters:



From last time

$$I_L = |I_1| + |I_2|$$

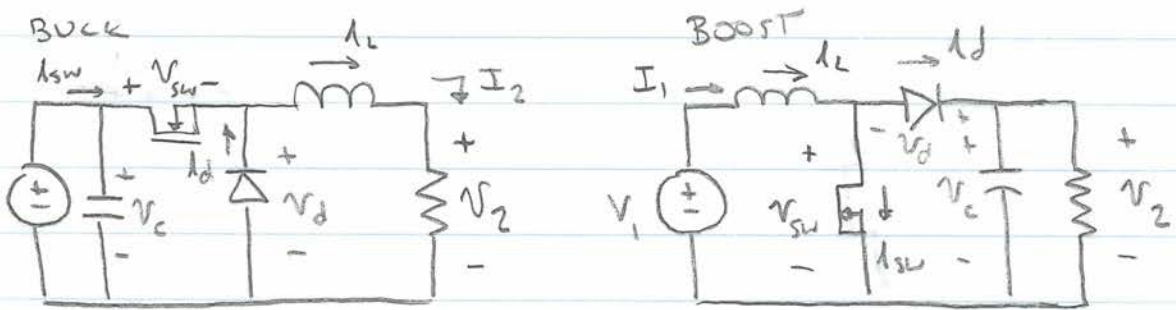
$$V_c = |V_1| + |V_2|$$

So we have

$$V_{sw,max} = V_{d,max} = V_c = |V_1| + |V_2|$$

$$I_{sw,max} = I_{d,max} = I_L = |I_1| + |I_2|$$

Lets look at direct converters (L, C big; $I_L \approx I_L, V_c \approx V_c$)



$$V_{sw,max} = V_{d,max} = V_c = V_1 \qquad V_{sw,max} = V_{d,max} = V_c = V_2$$

$$I_{sw,max} = I_{d,max} = I_L = I_2 \qquad I_{sw,max} = I_{d,max} = I_L = I_1$$

FOR THE DIRECT CONVERTER TYPES (L, C Big)

$$V_{sw,max} = V_{d,max} = V_c = \max(|V_1|, |V_2|)$$

$$I_{sw,max} = I_{d,max} = I_L = \max(|I_1|, |I_2|)$$

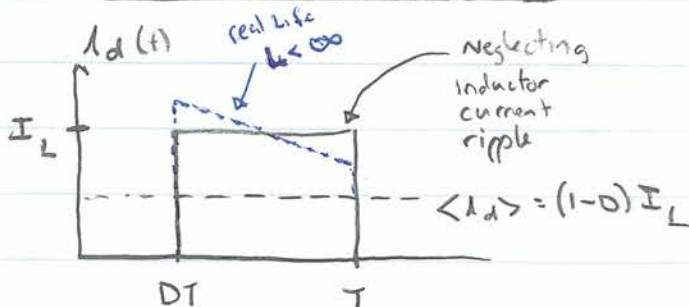
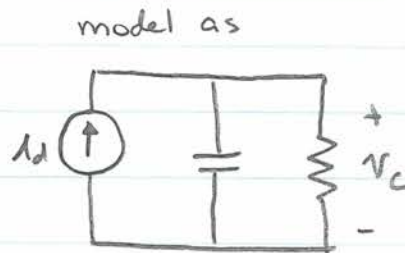
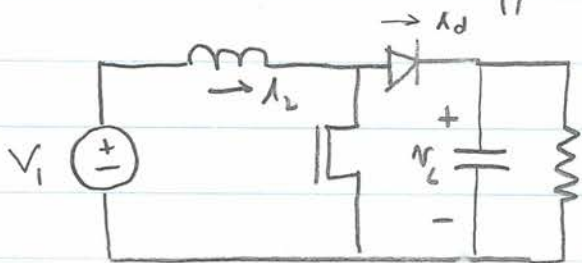
So, based on device and passive component stresses, we would choose a direct converter over an indirect converter whenever possible!

In practice, component selection does depend on ripple in many cases. Let's see how to approximately calculate ripple effects.

To calculate capacitor voltage ripple we

- ① Neglect ripple in inductor (assume $L \approx \infty$ so $\Delta I_{L,pp} \approx 0$)
- ② assume all current ripple goes into capacitor
- ③ Calculate voltage ripple
- ④ verify assumption afterwards

Ex: Boost Converter ripple

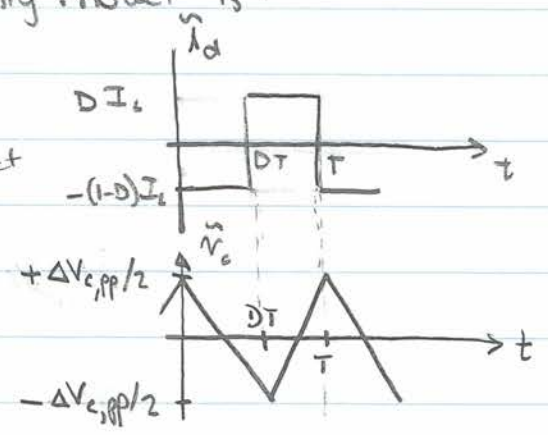
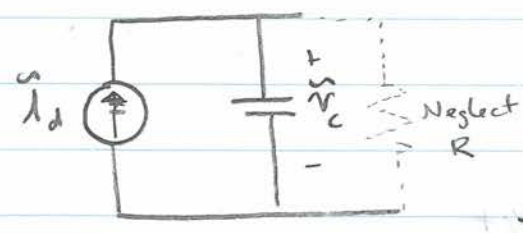


So we model the system assuming all ripple current component (\tilde{I}_d) goes into the capacitor, and the dc component $\langle I_d \rangle$ goes into the resistor. for this to be true $2\pi f_{sw} \gg \frac{1}{RC}$

6.334 Lecture

DC/DC Lecture #3

Under this assumption, a "ripple only" model is:



- ripple in v_c is triangular
- The average value of a Δ wave

is half-way between its peaks, so we get ripple between $\pm \Delta v_c / 2$

$$i = C \frac{\Delta v_c}{\Delta t}$$

$$i = C \frac{dv_c}{dt} \Rightarrow \Delta v_c = \frac{1}{C} \int i_c dt$$

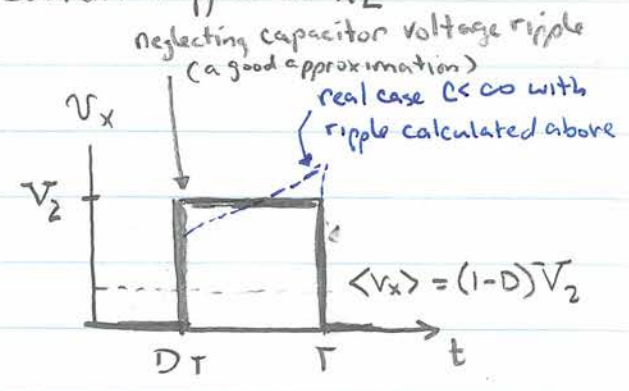
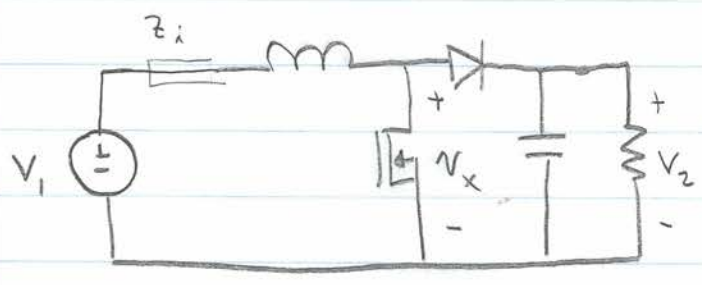
$$\Delta v_{c,pp} = \frac{D(1-D)I_L T}{C}$$

where $I_L = I$,

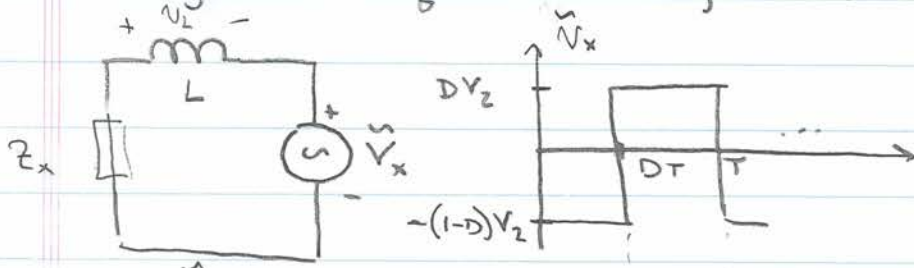
So to limit ripple below a specified value

$$C \geq \frac{D(1-D)I_L T}{\Delta v_{c,pp}}$$

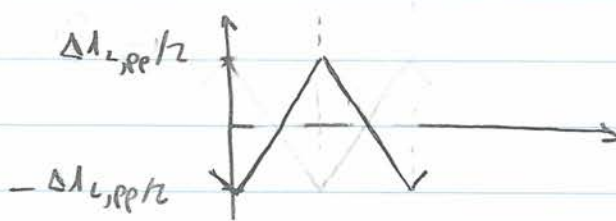
Similar calculations can be made for current ripple in i_L



modeling V_x with equivalent source, throw out dc components:



Neglecting Z_x , $V_L = -\tilde{V}_x$



$$V = L \frac{\Delta I}{\Delta t}$$

$$\Delta I_{L,pp} = \frac{D(1-D)V_2 T}{L}$$

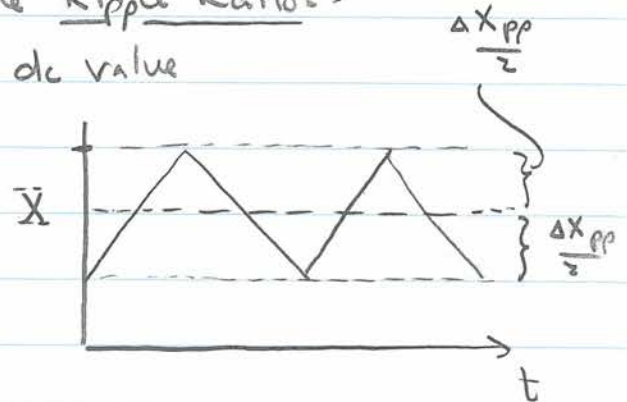
∴

$$L \geq \frac{D(1-D)V_2 T}{\Delta I_{L,pp}}$$

from these waveforms we can define Ripple Ratios:
ripple magnitude as a fraction of dc value

$$R_C \triangleq \frac{\Delta V_{C,pp}/2}{V_C}$$

$$R_L \triangleq \frac{\Delta I_{L,pp}/2}{I_L}$$



$$\therefore V_{C,pk} = V_C (1 + R_C) \quad I_{L,pk} = I_L (1 + R_L)$$

so from our previous results

$$L \geq \frac{D(1-D)T V_2}{2I_1 R_L}$$

$$C \geq \frac{D(1-D)I_1 T}{2V_c R_c}$$

Energy Storage is one metric for the minimum size of an energy storage component. What is required energy storage?

Capacitor $E_c = \frac{1}{2} C V_{c,plc}^2$

$$= \frac{1}{2} \frac{D(1-D)I_1 T}{2V_c R_c} [V_c(1+R_c)]^2$$

$$= \frac{1}{2} \frac{D(I_1 V_1) T}{2} \frac{(1+R_c)^2}{R_c}$$

$$E_c = \frac{DT P_o}{4} \frac{(1+R_c)^2}{R_c}$$

So energy storage increases with

- ① switching period
- ② output power
- ③ conversion ratio
- ④ smaller ripple spec.

Similar arguments for inductor $E_l = \frac{D \cdot P_o}{4 f_{sw}} \frac{(1+R_c)^2}{R_c}$

It can be shown that direct converters always require less energy storage (+ hence smaller components) than indirect converters.

We can also factor in ripple on our peak device stresses:

$$\text{DIRECT: } \begin{aligned} V_{c, pk} = V_{sw, pk} = V_{d, pk} &= \max(|V_1|, |V_2|)(1 + R_c) \\ I_{L, pk} = I_{sw, pk} = I_{d, pk} &= \max(|I_1|, |I_2|)(1 + R_L) \end{aligned}$$

$$\text{Indirect: } \begin{aligned} V_{c, pk} = V_{sw, pk} = V_{d, pk} &= (|V_1| + |V_2|)(1 + R_c) \\ I_{L, pk} = I_{sw, pk} = I_{d, pk} &= (|I_1| + |I_2|)(1 + R_L) \end{aligned}$$

IF time, define metric for switch size: switch stress parameter

$$\text{S.S.P.} \triangleq V_{sw, pk} \cdot I_{sw, pk}$$

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