

Receivers, Antennas, and Signals – 6.661

Problem Set No. 9

Issued: 4/08/03

Due: 4/17/03

Problem 9.1

a) Let $E_b = S/C = S/[B \log_2(1 + S/N_oB)] = E_bR/[(R/Q) \log_2(1 + RE_bQ/kTR)]$, so $Q = \log_2(1 + E_bQ/kT)$ and $2^Q - 1 = E_bQ/kT$. Therefore

$$E_b = kT(2^Q - 1)/Q = kT \text{ if } Q = 1. \text{ For small } Q, E_b \text{ diminishes only slightly}$$

b) Numerically, $Q = 2 \Rightarrow E_b = 1.5kT$; $Q = 0.3 \Rightarrow 0.77kT$, $Q = 0.1 \Rightarrow 0.72kT$.
 $Q = 0.001 \Rightarrow 0.69kT$. Lower values for E_b are generally unattainable in practice, so $E_b > \sim 0.7kT$ is a fairly hard limit, where T is the system noise temperature characterizing the channel, not the signal. Thus the desired range is: $0 < Q < 2$.

Problem 9.2

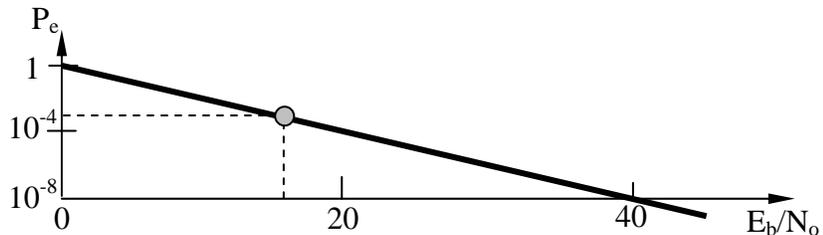
We can use up to 12 parity check bits. Using (4.5.6) in the text, where $K+R = 24$,

we find: $12 \geq \log_2[1 + \binom{24}{1} + \binom{24}{2} + \binom{24}{3} + \dots]$ where $\binom{24}{n} \equiv 24!/n!(24-n)!$;

Evaluating \Rightarrow

$$12 \geq \log_2[1 + 24 + 276 + 2024] = \log_2 2325, \text{ so } \boxed{\text{we can correct up to 3 errors.}}$$

Problem 9.3

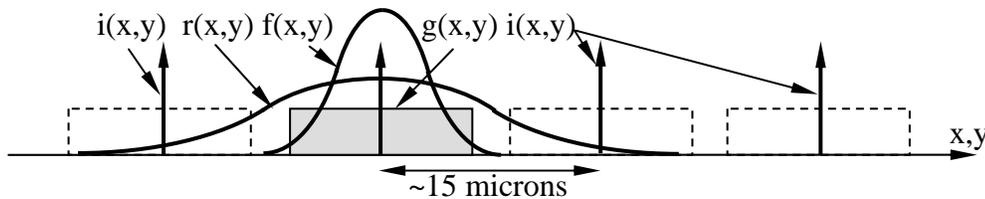


We may refer to (4.5.15) and surrounding text for help here. We lose 2.4dB in E_b/N_o because we have to speed up to make room for the parity bits, but we reduce the probability of block error from 10^{-3} to $Z = (1-10^{-3})^5 P_e^2 \times 7 \times 6/2!$. Since a block error would result on average in half the message bits being in error, the probability of message bit error would be $(2/7)Z$, or $\sim 6 \times 10^{-6}$. Thus P_e would be reduced by a factor of 168, corresponding to a change in E_b/N_o of $22.2/2 = 11.1$ dB (Note: $\log 168 = 22.2$ dB, and 40 dB change in E_b/N_o corresponds to 80 dB in P_e). The coding gain is the difference between the penalty of 2.4 dB and the reward of 11.1 dB, or $G_c \cong 11.1 - 2.4 = \boxed{8.7\text{dB}}$

Problem 9.4

b) The noise $n_1(t)$ can be considered to be the environmental noise, while $n_2(t)$ is post-smoothing and can include the sensor noise, photon shot noise, read-out noise, and quantization noise. Lumping the sensor noise with $n_1(t)$ is awkward because $n_1(t)$ precedes convolution.

a) Let x,y be coordinates referenced to the focal plane. We can consider $f(x,y)$ to correspond to the lens blurring function (arbitrarily chosen to be narrow) and $g(x,y)$ to be the focal plane response function of each CCD pixel; $i(x,y)$ represents the spacing of the pixels and $r(x,y)$ corresponds to the viewer's visual response function (assumed blurry). Thus we might have:



Problem 9.5

The received power must support the desired output SNR, which is $20 + 40$ dB ($\Rightarrow 10^6$). But for SSBSC, which performs the same as DSBSC, the necessary S/N_{out} is $\langle s^2(t) \rangle (P_c/2N_oW)$ where $N_o = kT_s/2$, $T_s = 4000K$, $k = 1.38 \times 10^{-23}$, and $W = 10^4$ Hz. $\langle s^2(t) \rangle = 0.5$ for pure sine waves at maximum amplitude, and 1 for square waves. The maximum average received power is then $P_c \langle s^2(t) \rangle = 2N_oW \times 10^6 = 2 \times 1.38 \times 10^{-23} \times 4000 \times 10^4 \times 10^6 = 1.1 \times 10^{-9}$ [W]. If we allow for a 70-dB path loss, then the average transmitter power is $\sim 1.1 \times 10^{-2}$ W, or ~ 10 milliwatts, which is reasonable.

Problem 9.6

a) Referring to Figure 4.7-10 and associated text, the FM threshold S/N for $\beta^* = 10$ is approximately 18 dB, so $P_c/BkT = 10^{1.8}$ and $P_r = P_c > 63 \times 2W(1 + \beta^*)kT = 63 \times 2 \times 10^4 (1+10) 1.38 \times 10^{-23} \times 4000 = 7.7 \times 10^{-13}$ W received and 7.7×10^{-6} W transmitted

b) The output SNR requirement is $20+40 = 60$ dB, where $S_{out}/N_{out} = P_c \langle s^2 \rangle / 3\beta^{*2} / 2N_oW = 10^6$, so $P_c \cong 10^6 \times 2 \times 1.38 \times 10^{-23} \times 4000 \times 10^4 / (0.5 \times 3 \times 100) = 7.36 \times 10^{-12}$ W, $P_t \cong 7.4 \times 10^{-3}$ W.

c) The requirements for P_t are a factor of ~ 9.6 greater than the FM threshold, so we could reduce β^* slightly to x , where $10^2/x^2 = 9.6$, so the new $x = \beta^*$ could be 3.23, but then the only margin left would be due to the fact that the FM threshold drops slightly with β^* .